An exciting hobby....for everyone

MAY 72 eyeryaa electronics



A POCKET GUIDE TO

CONSTRUCTIONAL **ELECTRONICS**

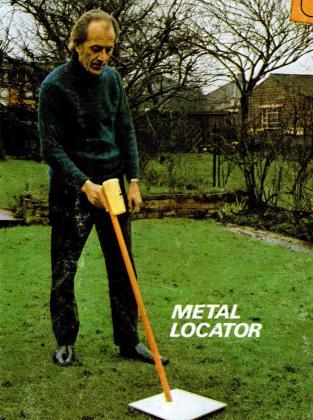
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West Africa 3/6d Sweden Kr. 3.00





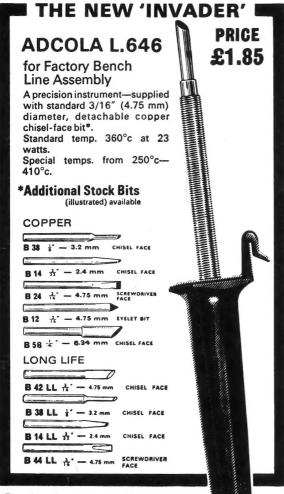
CONSTRUCTO OMPANION







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Don't take chances. We don't. All our ADCOLA Soldering Instruments are of impeccable quality. You can depend on ADCOLA day after day. That's why they're so popular. You get consistent good service...reliability...from our famous thermally controlled ADCOLA Element and the tough steel construction of this ideal production tool.

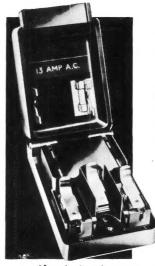


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ADCOLA PRODUCTS LTD.,

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SAFEBLOC



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Safe, quick and secure it connects 2-core and 3-core bare-ended flexible leads to the mains (A.C. only).

The concept was pioneered by Rendar, and introduced to the market 13 years ago.

Safebloc saves time. No need to fit a plug for tests. No danger, as no current can pass with the lid open.

Invaluable for testing and demonstrations in industry and shops, the work bench and the home.

Ask for Safebloc at your local stockist — or you can order it direct from the manufacturer.

If ordering by post, send cash with order.
PRICE £2.60+10p P.&P. EACH
Special bulk order wholesale and industrial rates on application



RENDAR

Rendar Instruments Ltd., Victoria Road, Burgess Hill, Sussex.Tel. Burgess Hill 2642

BARGAIN PRINTED CIRCUIT OFFER

Circuit Board with all holes drilled, 7½" × 5½" inc. central hole 1½" for speaker magnet and cut out for PP9 batt.; Rocker w/change switch and mounting bracket; 2 gang tuning capac.; 3 L.F.S., Osc. Coil, Ferrite rod with coils and holder, Potentiometer and knob; Circuit Booklet showing component values and positions. All for £1.75 (25p Post). Worth £5.

BATTERY CHARGER

5½" × 3" × 3" with fixing feet; 12V 2Amp. On-off Indicator, 2 yds. Mains and 2 yds. Battery Leads; Battery Clips. £1.50 (25p. Post).

PANEL METERS—70mm square. Minus 10A. to Plus 20A. D.C. £1 (15p Post); 2" (15p. Post).

STEREO AMPLIFIER Type SHV-2 x 3 watts

Fully built. Separate vol., bass and treble controls each channel; 12 × 4½ × 6in high. EZ80, ECC83, 2 × ECL86 valves. O.P. trans. for 3-ohm speakers. Double wound mains trans. Suitable for crystal, magnet cartridge, tuner, etc. 200-250V. A.C. mains. 27, 50p P. & P.



30° 0'4)

MONO GRAM CHASSIS 3 WATT

3 Wave band long-med.-short, Gram., 200-250V. A.C. Ferrite aerial. Chassis 13 × 7 × 5in. Dial 13 × 4in. Double wound mains transformer 5 valves ECH81, EF89, EEC81, EL84, EZ80. Price \$10-63. (37p P. & P.) Output trans. for 3-ohm speaker.
Some alightly tarnished at \$10 carr. pd.

MAINS TRANSFORMERS (240-250V input)

Postage in brackets. 6:3V at 2½A. 40p (15p) 280-0-280V 60MA, 6:3V 2½A, 6:3V 700mA £1 (27p) 250V at 50mA and 6:3V at 1½A. 50p (20p) 22V at 1A, 6:3V at 2½A and 250V at 50mA. 75p (25p) 90V at 20mA and 1:4V at 250mA. 50p (15p) Deduct 10 per cent from total bill for more than one transformer.

GLADSTONE RADIO

66 ELMS ROAD, ALDERSHOT, HANTS.

(2 mins. from Station and Buses). FULL GUARANTEE. Aldershot 22240. CLOSED WEDNESDAY. S.A.E. for enquiries please.

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HOME RADIO (COMPONENTS) LTD., Dept. EE, 234-240 London Road, Mitcham, Surrey CR4 3HD.



ystem sounds great itication sounds

Viscount III Audio Suite complete £49

20 14 + 14W per channel 40Hz (40kHz ± 3dB Total distortion at watts at 1kHz -0.1%.

This is real value for money! We have designed 3 systems and the heart of them all is the Viscount III amplifier. A unit of great eye appeal with teak finished cabinet. It is available in 2 versions—Rion for ceramic carrridges, and RIOI for magnetic and ceramic. FET's (Field effect transistors) like top priced units. FET's give you more of the signal you want and almost none of the hiss you don't. Both units have output Filters and tone controls give a wide range are incorporated on the input stages, just sockets for headphones and tape recorder. bass and treble adjustment. ŏ

For all systems we have chosen the famous Garrard SP25 Mk. III deck which comes complete with teak finished plinth and perspex cover.

The exclusive Duo loudspeaker systems Duo Il's for the smaller room or the big substantial cabinets. There's a choice of the are incomparable for quality within their price range. Large speakers in extremely Duo III's for real bass response.

PRICES SYSTEM I

£22.00+90p p&p £14.00+£2 p&p £23.00+£1.50 Viscount III R101 amplifier £. 2 × Duo Type II speakers £ Garrard SP25 Mk. III with MAG. cartridge plinth and cover £

3 ohms. Simulated Teak cabinet.

£14 pair+£2 p&p.

SPEAKERS Duo Type II

£59·00 Total

for only £52+£3.50 p&p Available complete

£22.00+90p p&p £32.00+£3 p&p £23.00+£1.50 2 × Duo Type III speakers £ Garrard SP25 Mk. III with MAG. cartridge, plinth and cover SYSTEM 2 Viscount R101 amplifier

00.773 Available complete Total

2 × Duo Type II speakers, pair £17·00+90p p&p Garrard SP2 All III with CER. SYSTEM 3 Viscount III Amplifier R100 diamond cartridge, plinth and cover

for only £49+£3.50 p&p Available complete

Size approx. $17" \times 10\frac{2}{3}" \times 6\frac{2}{3}"$. Drive unit $13" \times 8"$ with parasitic tweeter. Max. power 10 warts, Drive unit 134" x 84" with H.F. speaker. Max. power 20 watts at 3 ohms. Freq. range 20Hz to 20kHz. Teak veneer cabinet. £32 pair +£3 p&p. Duo Type III. Size approx. 23½" × I1½" × 9½".

Goods not despatched outside U.K.

21c High Street, Acton, London W3 6NG, 323 Edgware Road, London, W2. Mail orders to Acton. Terms C.W.O. All enquiries S.A.E. Radio and TV Components (Acton) Ltd.,

£21.00+£1.50 for only £69+£4 p&p £52.00 Total

SPECIFICATION R101
If wates per channel into 3 to 4 ohms. Total distortion @ 10W @ 1kHz 0-1%. P.U.J (for ceramic carridges) 150M° tinto 3 Meg. P.U.2 (for magnetic carridges) 150M° tinto 3 Meg. P.U.2 (for magnetic carridges) 150M° tinto 3 Meg. P.U.2 (for magnetic carridges) 4mV @ 1kHz into 47K. equalised within ±1dB R.L.A.A. Radio 150mV into 20K (Sensitivities you at till power). The out facilities: headphone socket, power out 250mV per channel. The fore controls and filter than testifiers. Table Both 2 Meg. Gold. Bass filter: 6dB per octave cut. Treble (inter: 12dB per octave, 5/gnd in onise radio (all controls at max) R101—P.U.1 and radio—65dB. P.U.2—58dB. R105 same R10 but P.U.2 (for crystal carridges) 450mV into 3 Meg. Cross talk better than 35dB on all inputs. Overload characteristics better than 26dB on all inputs. Size approx 13g* x y* x 3g*.

FOR RAPID SERVICE

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CAPACITORS

TRANS	STO	RS		_
AC127 AC127 AC127 AC176 AC176 AC187 AC189 AD1199 AD1160 AF118 AF124 AF125 AF126 AF127 AF127 BC108 BC108 BC147 BC109 BC147 BC158 BC158 BC159 BD131 BD132	17p 18p 22p 22p 23p 47p 62p 45p 19p 67p 45p 11p 11p 11p 11p 12p 14p 175p	BFX29 BFX84 BFX88 BFY50 BFY51 MAT101 MAT120 MAT121 OC28 OC35 OC45 OC72 OC72 OC72 OC201 OC71 OC73 OC201 OC71 UT46 2N696 2N29269 2N29267 2N29267 2N29269 2N29269	38p 25p 30p 21p 21p 22p 22p 22p 22p 48p 112p 112p 112p 23p 23p 23p 23p 12p 112p 1	
BF115 BF178	25p 32p	2N3054 2N3055	60p 72p	
BF115 BF178 BF179	25p 32p 56p	2N3054 2N3055 2N3702	60p 72p 15p	
BF180 BF181 BF184 BF185	30p 32p 30p 32p	2N3703 2N3704 2N3705 2N3706	14p 15p 14p 14p	
BF194 BF195	14p 14p	2N3711	14p	

DIODES

AAII9	Пp	OA202	10p
OA47	71p	BY100	15p
QA90	7‡p 6p	BY127	22 p
OA9I	6р	BYZ12	22 i p

2 N 4058 2N5459

60p

ZENER DIODES From 2 to 33 volts. 400mW, 15p; 1.5W, 221p

SILICON BRIDGE RECTIFIERS

FUSES AND HOLDERS

liin glass—2ip 60, 100, 150, 250, 500, 750mA: 1, 125, 15, 2, 2-5, 3, 5, 7-5, 10, 15 amp. iin glass—2ip 100, 250, 300mA; 1, 2-5 amp. Anti-eurge 1;in—8p 200, 500, 750, 850mA; 1, 1-5, 2, 3 Anti-surge 20mm—5p 80, 125, 200, 315, 400, 500, 630, 800mA; 1, 2 amp.

PANEL FUSEHOLDERS

For Itin fuses For 20mm fuses

CONTROLS, Log. or Lin. Single, less switch, 15p Single, D.P. switch, 24p Tandem, less switch, 40p 5k Ω , 10k Ω , 25k Ω , 50k Ω , 100k Ω , 250k Ω , 500k Ω , 1M Ω , 2M Ω

RESISTORS

Carbon All 5%, high-stability, E12 values. ½W, 1½p; IW, 4p; 2W, 6p Wire-wound 5W, 10p; 10W, 12p

ELEC	TRO	LYTIC	cs		
1μF 2μF 4μF 8μF 16μF	450V 500V 350V 450V 450V	19p 20p 14p 16p 17p	1,000µF 1,000µF 2,000µF 2,000µF 2,500µF	25V 50V 25V 50V 25V	27p 39p 36p 53p 45p
25μF 32μF 50μF 100μF	50V 450V 50V 25V 50V	8p 24p 10p 10p 10p	2,500µF 3,000µF 5,000µF 5,000µF 8–8µF	50V 25V 25V 50V 450V	60p 48p 55p 98p 18p
250μF 250μF 500μF 500μF	25 V 50 V 25 V 50 V	12p 17p 18p 25p	8-16μF 16-16μF 16-32μF 32-32μF 50-50μF	450V 450V 450V 450V 350V	20p 27p 63p 49p 38p

A11 L4 1	MIUKE	ELECT	RULI	1103
IμF	25V	IOμF	64V	
2 5uF	64V	16µF	40V	
4µF	40V	25µF	25V	7
5μF	64V	30µF	15V	
8µF	15V	50µF	15V	
8µF	40V	100µF	15V	
10µF	15∨			•

ALUMINIUM BOXES with lids and screws

Type	Length	Width	Depth	Price
GB7*	2∄in	5±in	1±in	38p
GB8*	4in	4in	l√in	38p
GB9*	4in	2∄in	l∮in	38p
GBI0*	4in	5≟in	I∛in	44p
GBII	4in	2√in	2in	38 _P
GB12	3in	Žin	lin	33p
GB13	6in	4in	2in	52p
GB14	7in	· 5in	2∔in	63p
GB15	8in	6in	3in	8lp
GB16	l Oin	7in	3in	92p
*	These sizes	fit standard	Veroboa	ards .

Size	0-1 matrix	0·15 matrix
2½in × 3≵in 2½in × 5in	22p	16p
2-in × 5in	24p	25p
3∄in × 3∄in	24p	25p
3∦in × 5in	27p	29p
l7in × 2-jin	75p	57p
l7in × 3≩in	£I	75p
Pins-both	sizes: packet	of 36. I&p

VARIABLE POWER SUPPLY

Input: 240V, a.c.
Output: Switched 3, 4-5, 6, 7-5, 9, 12 volts d.c. at 500mA

CASSETTE OWNERS!

For Philips and similar cassette recorders. PU12 Power unit for connection to 12V + or — E car electrical systems, giving 7½V, stabilised output.

PP75 Mains power supply, output £1.95 Both units are complete with cable and 5 pin D.I.N. plug

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B.A.F. wadding, 18in wide, 1in thick. The ideal lining for speaker enclosures. 25p per

MISCELLANEOUS ITEMS

B9A valve bases, 2p 5kΩ edge control, fits most small, imported radios, 7p 20Ω volume control for 3Ω speakers, 20p Antex CN240, 15W miniature soldering iron, £170 alve and Transistor Data book, 9th edition,

75p
Transistor equivalent book, BPI, 40p

LOW-OHM RESISTORS

 $2\frac{1}{2}$ watt wire-wound. 1Ω , 1.8Ω , 2.7Ω , 3.3Ω , 3.9Ω , 4.7Ω , 5.6Ω , 6.8Ω , 8.2Ω



CAPA	CIIO	K S		0.003µF	500V	Cer.	5p
2-2pF	500V	S/M	7 ½ p	0 0033μF	125V	P.S.	6p
3-3pF	500V	S/M	7 P 7 P	0·0033μF	500∨	Poly.	6p
5pF	500V	S/M	7 1 P	$0.0033 \mu F$	1,000V	MDC	6p
10pF	125V	P.S.	5p	0·0036μF	500V	S/M	15p
10pF	500∨	S/M	71p	0 0047μF	125∨	P.S.	9p
15pF	125V	P.S.	5p	0·0047µF	500∨	Poly.	6р
15pF	500V	Cer.	4p	0.0047µF	500V	S/M	20p
18pF	500V	S/M	71p 5p	0·0047µF	1,000∨	MDC	6p
22pF	125V	P.S.	5p	0·005μF	100	Mylar	3р
22pF	500V	S/M	7 1 P	0·005μF	500∨	Cer.	5p
25pF	500V	S/M	71p	0·0068μF	125V	P.S.	101p
27pF	500V	Cer.	4p	0·0068μF	500V	S/M	30p
33pF	125V	P.S.	5p	0·0068μF	500V	Poly.	6p
33pF	500V	S/M	7 } P	0·0082μF	125V	P.S.	101p
39pF	500V	S/M	71P 71P 50	0 0082μF	500V	S/M	30p
47pF	125V	P.S.	3p	0.01 µF	120	Disc	4p
47 pF	500∨ 500∨	Cer. S/M	4p	0.01μF	125V	P.S.	101p
50pF	500V	S/M	7.1P	0.01μF	160V	Poly.	4p
56pF	1257	P.S.	7 P	0.01μF	250V 400V	M.F.	3p
68pF 68pF	500V	S/M	,5p	0 01μF	500V	Poly.	3p
75pF	500V	S/M	71p 71p 71p 71p	0.01μF	500 V	Cer. S/M	35p
82pF	500 V	S/M	/1P	0.01μE	600V	MDC	30p
100pF	1257	D C	'IP	0·01μF 0·01μF	1,0007	MDC	7p 9p
100pF	500V	P.S. S/M	5p 7∳p	0.015μF	1600	Poly.	3p
100pF	500 V	Cer.	′1p	0.015µF	400V	Poly.	3p
120pF	500V	S/M	7 1 P	0.0345	100V	Mylar	3p
150pF	125V	P.S.	5p	0·02μF 0·022μF	187	Disc	5p
150pF	500V	S/M	71P	0.022µF	250V	M.F.	3p
150pF	500V	Cer.	155	0.022µF	400V	Poly	15
180pF	500V	S/M	5p 71p	0.022µF	600V	Poly. MDC	3p 71p
200pF	500V	S/M	7 1 P	0.022µF	1,0000	MDC	9p
220pF	125V	P.S.	Šp	0·033μF	250V	M.F.	4p
220pF 220pF	500V	Cer.	5p	0.033µF	400V	Poly.	4p
250pF	500V	S/M	8p	0.047µF	120	Disc	6р
270pF	500V	Cer.	5p	0.047µF	160V	Poly.	3p
300pF	500V	S/M	8p	0.047µF	250V	M.F.	3р
330pF	125V	P.S.	5p	0.047µF	400V	Poly.	4p
330pF	500V	S/M	8p	0·047µF	600V	MDC	8p
390pF	500∨	S/M	8р	0·047μF	1,000V	MDC	10p
470pF	125V	P.S.	5p	0·1μF	30∨	Disc	6р
470pF	750∨	Disc	5p	0·1μF	250∨	M.F.	4p
500pF	500V	S/M	8р	0·1 μF	400V	Poly.	5p
560pF	500V	S/M	8р	0·1μF	600V	MDC	I0p
680pF	125V	P.S.	6р	0·1μF_	1,000V 250V	MDC	13p
680pF	500V	S/M	8p	0·15μF	250V	M.F.	5p
820pF	500V	S/M	8p	0 22μF	160V	Poly.	6p
0.001μF	100V	Mylar	Зp	0·22μF	250V	M.É.	5p
0.001µF	125V	P.S.	6р	0·22μF	400V	Foil	10p
0·001μF	400V	Poly.	3p	0·22μF	1,000V	MDC	I5p
0.001µF	500V	S/M	10p	0·33μF	250V	M.F.	8p
0 00 I μF	500V	Cer.	5p	0·47μF	250V	Foil	.8p
0·001μF	1,000V	MDC	6р	0·47μF	400V	Foil	15p
0.0015μF	400V	Poly .	3p 10p	0·47μF	1,000V 250V	MDC	20p
0.0015μF	500∨ 500∨	S/M	IUP	I·0μF	250 V	M.F.	15p
0·0015μF 0·0018μF	500V	Cer. S/M	5р 10р	Note:			
0.0018μF	1007	Mylar	3p	S/M	lver mica	10/	1
0.002µF	500V	Cer.	5p	P S = D	lystyren	- 220%	tol
0.0022µF	125	PS.	6p	MDC	a.c. ratin	v = 300	v
0 0022µF	500V	P.S. S/M	10p		fullard m		
0.0022µF	1.000	MDC	6p	Cer. = c			
	.,						

0.0027µF

500V S/M

PLUGS

Car aerial
Co-axial
D.I.N. 2 pin (speaker)
D.I.N. 3 pin
D.I.N. 4 pin
D.I.N. 5 pin, 180
D.I.N. 5 pin, 240°
D.I.N. 6 pin
lack. 24mm unscreene D.I.N. 6 pin
Jack, 2½mm unscreened
Jack, 2½mm screened
Jack, 3½mm unscreened
Jack, 3½mm screened
Jack, ½in unscreened
Jack, ½in screened Jack, stereo, unscreened Jack, stereo, unscreened Jack, stereo, screened Phono, plastic top Phono, plated metal Phono, fitted 4 ft lead Wander, red or black Banana 4mm, red or black

LINE SOCKETS Car aerial
Co-axial
D.I.N. 2 pin (speaker)
D.I.N. 3 pin
D.I.N. 5 pin, 180
D.I.N. 5 pin, 240 Jack, 3½mm Jack, ½in screened Jack, stereo, screene Phono, plated metal



SOCKETS

Car aerial
Co-axial, surface
Co-axial, flush
D.I.N. 2 pin (speaker)
D.I.N. 3 pin
D.I.N. 5 pin, 180°
D.I.N. 5 pin, 240°

D.I.N. 5 pin, 240°
Jack, 24mm
Jack, 34mm
Jack, 34mm
Jack, 41n unswitched
Jack, 41n switched
Jack, stereo, switched
Phono, 5 ingle
Phono, 2 on a strip
Phono, 3 on a strip
Phono, 4 on a strip
Wander, single, red or black
Wander, twin strip
Banana 4mm red, or black

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Size 1 cartons all at 25p each in 40/60, 60/40, or Savbit alloys in 7 gauges.

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Model 3A. Strips insulation from cable or flex without nicking wire.4 different settings, 4&6 BAspanner ends, ground cutting edges Price 32p. Also available. **NEW!** Also available, de luxe Model 8. Price 58p.

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Would you believe it, some people still house their constructed projects in 'tatty' looking cases 'bashed out' from aluminium, tin plate, etc., or get in 'tatty' looking cases 'bashed out' involved in complicated carpentry.

No need for this any more !!!

We can supply cases in kit form to house most of your Everyday Electronics projects which will give them a touch of professionalism and a modern appearance with a mirror-like finish. Our cases, which are fabricated from Perspex, Oroglass and Acrylic Sheet (both in dense and transparent bright colours), are supplied in kit form, their construction being quick and easy and involving the minimum amount of work. Included with the kits are comprehensive assembly details and diagrams. Take the hard work out of making your cases and go MODERN with KASPEX.

ASTRON £1:40 DARKROOM TIMER £1:30 RAIN ALARM £0.90 SOIL MOISTURE METER £1.00*
REMOTE TEMP. COMP? £1.30*
BABY ALARM (pair) £1.50

Mail Order only-Cash with order.

State diameter of meter cut-out required when ordering. All prices include post and packing.

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need components of professional quality. Our Catalogue shows practically the entire range marketed by R.S. Components Ltd. ALL of these components, used extensively by industry, are now made readily available to

MAIL ORDER ELECTRONIC COMPONENTS P.O. BOX No. I, LLANTWIT MAJOR, GLAMORGAN CF6 9YN

Everyday Electronics, May 1972



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MO	1/2W	2 %	10 Ω-IM Ω	E24	4	3 · 5	3
ww	IW	$10\% \pm 1/20 \Omega$	0 · 22 Ω-3 · 9 Ω	EI2	7	7	6
ww	3W	5 %	12 Ω-10K Ω	EI2	7	7	6
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Armstrong 525 (Teak cased)	£68 · 50
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(Teak cased)	£78 · 50
Leak Delta 75	£130 · 95
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Philips RH 882 (+ cass head)	£72 00
Philips RH702	£82.50
Teleton 2100	£29.95
	£105 · 00
Rogers R/brook (Teak)	£80.00
Rogers R/brook (Chassis)	£74.50

SPEAKERS

Please add £1 .25 P. & P.	per pair
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Wharfedale Denton 2	£29.00
Wharfedale Linton 2	£37·00
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Garrard SP25; 2025TC;	3000;
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Please add 10p for P. & P.	
Goldring G850	£3 · 65
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AF117	25p	B8X20	16p	NKT713	29p	ZTX304	27p	2N3704	11p
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AF124	25p	BSY27	20p	OA10	25p	ZTX320	30p	2N3706	9p
AF126	17p	BSY29	25p	OA47	8p	ZTX330	18p	2N3707	11p
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AF186	40n	BY100	15p	OA73	8p	ZTX501	16p	2N3709	9p
AF239	36p	BY127	15p	OA79	8p		20p	2N3710	9p
AF279	47p	BYZ10	35p	OA81	8p	ZTX503	17p	2N3711	9p
ASY26	25p	BYZ12	a00	OA85	8p	ZTX504	40n	2N3819	35p
ASY27	30n	BYZ13	20p	OA90	8p	IN914	7p	2N3820	60p
ASY28	22p	BZY88		OA91	8p	IN4001	7n	2N3826	30p
ASY29	30p	C3V3	15p	OA95	8n	IN4002	7p	2N4058	15p
ASZ21	37p	C3V6	15p	OA200	10p	IN 4003	100	2N4060	12p
BC107	10p	C3V9	15p	OA202	10n	1N4004		2N4061	12p
BC108	10p	C4V3	15p	OC19	37p	IN4005	12n	2N4062	12p
BC109	10p	C4V7	15p	OC20	97p	IN 4006	Top	2N4289	15p
BC147	10p	C5V1	15p	OC22	47p	2N5756	95p	2N4871	40p
BC148	9p	C5V6	15p	OC23	60p	IN4007	20p	2N5245	45p
BC149	10p	C6V2	15p	OC24	60p	IN4148	7p	40250	55p
BC158	11p	C6V8	15p	OC25	37p	2G302	тар	40309	33p
BC167	11p	C7V5	15p	OC26	33p	2G371	15p	40310	45p
BC168	10p	C8V2	15p	OC28	60p	2G374	25p	40312	48p
BC169	11p	C9V1	15p	OC29	60p	2N174	80p	40320	47p
BC169C	15p	C10	15p	OC35	50p	2N385A/	50p	40360	43p
BC182	10p	C11 .	15p	OC36	63p	2N388A	50n	40361	47p
BC182L	10p	C12	15p	OC41	25p	2N404	23p	40362	55p
BC183	9p	C13	15p	OC42	30p	2N696	15p	40406	56p
BC183L	10p	C15	15p	OC44	15p	2N697	17p	40407	39p
BC184	13p	C16	15p	OC45	12p	2N698	30p	40408	51p
BC184L	12p	C18	15p	OC71	12p	2N706	10p	40409	54p
BC212	12p	C20	15p	OC72	12p	2N706A	12p	40410	62p
BC212L	12p	C22	15p	OC75	23p	2N708	16p	40468A	35p
BCY30	25p	C24	15p	OC76	25p	2N711	37p	40600	58p
BCY 31	48p	C27	15p	OC77	40p	2N711A	37p	40601	55p
BCY32	50p	C30	15p	OC81	20p	2N911	50p	40602	40p
BCY33	20p	CR1/0510	40p	OC81D	20p	2N914	20p	40603	49p
BCY34	25p	CR1/4010	60p	OC81Z	55p	2N918	42p	40486	95p
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NEW "SEW" CLEAR PLASTIC



1mA

METERS TYPE SW.100 100 x 80 mm

100 X 00	
20V. D.C	£3·10
50V. D.C	£8·10
300V. D.C.	£8·10
1 amp. D.C.	£3.10
5 amp. D.C.	£3·10
300V. A.C.	£3·10
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METERS TYPE S-80 80 mm. square fronts 50μA **£3·20** 50-0-50μA **£3·10**

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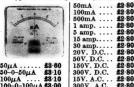


50V. D.C	£2·60
300V. D.C.	\$2.60
1 amp. D.C.	£2·60
5 amp. D.C.	£2·60
300V. A.C.	£2·60
VII Meter	42.27

"SEW" CLEAR PLASTIC METERS

500μA 1mA . 20V. I D.C...

Type MR.85P. 41in. × 42in. fronts.



50μA \$3·60
50-0-50µA \$8.10
100μA \$3·10
100-0-100µA #8 00
200μA £3 00
500μA £2.90
500-0-500μA £2·80
1mA \$2.80
1-0-1mA £2.80
5mA £2.80
10mA £2.80

500mA \$2.8
1 amp 23.80
5 amp £2-8
15 amp \$2.8
30 amp £2.9
20V. D.C \$2.8
50V. D.C £2.8
150V. D.C. 42.8
300V. D.C. #2.8
15V. A.C 42-8
300V. A.C. #2-8
8 Meter 1mA 22.8
VU Meter 23 6
1 amp. A.C. #2-8
5 amp. A.C. #2.8
10 amp. A.C. *22-8
20 amp. A.C. *22.8
30 amp. A.C. *22-8
in, square fronts.

Type MR.52P. 22in. square fronts.			
50μA	10V. D.C		
10mA \$2.00 50mA \$2.00 100mA \$2.00 5000mA \$2.00 1 amp \$2.00 5 amp \$2.00	VU Meter		

Type MR.65P. 3	in. × 3 in. fronts.
50μA £3·37	10V. D.C £2.20
50-0-50µA \$2.75	20V. D.C £2.20
100µA \$2.75	50V. D.C 42-20
100-0-100µA\$2.65	150V.D.C £2.20
200μA \$2.65	300V. D.C. \$2.20
500μA £2·40	15V. A.C \$2.30
500-0-500μA £2·20	50V. A.C \$2.30
1mA £2 20	150V. A.C. £2.30
5mA #2.20	300V. A.C. 42-80
10mA £2·20	500V. A.C. #2-30
50mA £2.20	8 Meter 1mA \$2.37
100mA £2.20	VU Meter 23.37
500mA £2⋅20	50mA A.C. *22.20
1 amp £2 20	100mA A.C. *£2.20
5 amp £2.20	200mA A.C. *\$2.20
10 amp £2.20	500mA A.C. \$2.20
15 amp £2.20	1 amp. A.C. * \$2.20
20 amp £2.20	5 amp. A.C.* £2.20
30 amp £2.30	10 amp. A.C. *22.20
50 amp 42.50	20 amp. A.C. *22.20
5V. D.C £2.20	30 amp. A.C. *42.20

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SEW EDUCATIONAL METERS



Type ED.107. Size overall 100mm × 90mm × 108mm. A new range of high quality moving coil instruments ideal

instruments ideal for school experiments and other bench applications. 3" mirror scale. The movement is easily accessible to the constrate internal working. Available actions are appeared to the constraint of t in the following ranges:

50μA 25·00	20V d.c £4.40
100µA £4.65	50V d.c \$4.40
1mA 24.40	300V d.c \$4-40
50-0-50μA £4-65	
1-0-1mA 24-40	Dual range
1A d.c £4-40	500mA/5A d.c.24-65
5A d.c £4.40	5V/50V d.c 24-85
10V d.c #4-40	

Type MR.38P. 1 21/32in. square fronts.

50μA. ±2:10 50μA. ±2:10 50μA. ±2:10 50μA. ±2:10 50μA. ±1:60 50μA. ±1:60 50μA. ±1:60 50μA. ±1:60 100μA. ±1:90 100μA. ±1:90 100μA. ±1:75 500μA. ±1:65 500ν ±1:60	y amount no more and amount of	200mm
\$\frac{\psi}{\psi} \frac{\psi}{\psi} \p	Landra	300mA \$1.60
150µA 11.60	The state of the s	
1 amp. 21.60 5 amp. 41.60 10 amp. 41.60 10 amp. 41.60 10 V. D.C. 41.60	~ V ~	
2 amp. 41.60	\$2,000 miles	
5 amp. 41.60 50μA 41.90 10V D.C. 41.60 100μA 41.90 10V D.C. 41.60 100-0-100μA 41.75 10V D.C. 41.60 200μA 41.75 10V D.C. 41.60 500-0-500μA 41.60 10V D.C. 41.60 500-0-500μA 41.60 10V D.C. 41.60 100-0-100μA 41.60 100V D.C. 41.60 100-0-100μA 41.60 100V D.C. 41.60 100-0-100μA 41.60 100V D.C. 41.60 100-0-100μA 41.60 10V D.C. 41.70 100mA 41.60 100V D.C. 41	200	
10 amp 21-60	- Tarr	
50μA \$2:10 3V. D.C. \$1:60 500-650μA \$1:90 10V. D.C. \$1:60 100μA \$1:75 10V. D.C. \$1:60 100-0-100μA \$1:75 20V. D.C. \$1:60 500μA \$1:85 150V. D.C. \$1:60 500μA \$1:85 150V. D.C. \$1:60 500μA \$1:85 150V. D.C. \$1:60 10μA \$1:80 300V. D.C. \$1:60 10μA \$1:60 100V. D.C. \$1:60 10μA \$1:60 100V. D.C. \$1:60 10μA \$1:60 10V. A.C. \$1:70 10μA \$1:60 150V. A.C. \$1:70 50μA \$1:60 500V. A.C. \$1:70 50μA \$1:60 50µA \$1:60	Albert Committee	
50μA \$2:10 3V. D.C. \$1:60 500-650μA \$1:90 10V. D.C. \$1:60 100μA \$1:75 10V. D.C. \$1:60 100-0-100μA \$1:75 20V. D.C. \$1:60 500μA \$1:85 150V. D.C. \$1:60 500μA \$1:85 150V. D.C. \$1:60 500μA \$1:85 150V. D.C. \$1:60 10μA \$1:80 300V. D.C. \$1:60 10μA \$1:60 100V. D.C. \$1:60 10μA \$1:60 100V. D.C. \$1:60 10μA \$1:60 10V. A.C. \$1:70 10μA \$1:60 150V. A.C. \$1:70 50μA \$1:60 500V. A.C. \$1:70 50μA \$1:60 50µA \$1:60	The state of the s	10 amp £1.60
50 ⁻ 0-50μΛ ±1:90 10V. D.C. ±1:60 100μΛ ±1:90 15V. D.C. ±1:60 100-0-100μΛ±1:75 20V. D.C. ±1:60 500μΛ ±1:75 500μΛ ±1:65 150V. D.C. ±1:60 500-0-500μΛ±1:60 300V. D.C. ±1:60 1mΛ ±1:60 500V. D.C. ±1:60 2mΛ ±1:60 150V. D.C. ±1:60 2mΛ ±1:60 150V. Λ.C. ±1:70 10mΛ ±1:60 150V. Λ.C. ±1:70 10mΛ ±1:60 150V. Λ.C. ±1:70 50mΛ ±1:60 500V. Λ.C. ±1:70 50	50µA £2·10	3V. D.C #1 60
100µA 21.90 15V. D.C. 21.60		
100-0-100µA\$1.75 200µA \$1.75 500µA \$1.86 500-0-500µA\$1.60 300V. D.C. \$1.60 500-0-500µA\$1.60 300V. D.C. \$1.60 1mA \$1.60 500V. D.C. \$1.60 2mA \$1.60 500V. D.C. \$1.60 2mA \$1.60 500V. D.C. \$1.60 500V. A.C. \$1.70 10mA \$1.60 50V. A.C. \$1.70 50mA \$1.60 500V. A.C. \$1.70		
200µA	100 0 100 4 81 88	
500µA \$1.465 150V. D.C. \$1.460 150V. D.C. \$1.460 150V. D.C. \$1.460 110A \$1.460 150V. D.C. \$1.460 110A \$1.460 150V. D.C. \$1.460 150V. A.C. \$1.70 150MA \$1.460 150WA \$1.460 150W		
500-0-500µA\$1:60 300V. D.C. \$1:60 1mA \$1:60 500V. D.C. \$1:60 1-0-1mA \$1:60 750V. D.C. \$1:60 2mA \$1:60 15V. A.C. \$1:70 10mA \$1:60 150V. A.C. \$1:70 20mA \$1:60 300V. A.C. \$1:70 50mA \$1:60 500V. A.C. \$1:70 100mA \$1:60 800V. A.C. \$1:70 100mA \$1:60 800V. A.C. \$1:70 50mA \$1:60 800V. A.C. \$1:70 500V. A.C. \$1:70 50mA \$1:60 800V. A.C. \$1:70 500V. A.C. \$1:		
1mA	500μA £1·65	150V. D.C. £1.60
1mA	500-0-500μA £1 60	300V. D.C. #1-60
1-0-1mA 41-60 750V D.C. 41-60 2mA 41-60 15V A.C. 41-70 5mA 21-60 50V A.C. 41-70 20mA 11-60 150V A.C. 41-70 20mA 41-60 300V A.C. 41-70 50mA 41-60 500V A.C. 41-70 100mA 41-60 8 Meter Ima 41-70		
2mA \$1.60 15V. A.C. \$1.70 5mA \$1.60 50V. A.C. \$1.70 10mA \$1.60 156V. A.C. \$1.70 20mA \$1.60 300V. A.C. \$1.70 50mA \$1.60 500V. A.C. \$1.70 100mA \$1.60 8 Meter ImA \$1.70		
5mA £1.60 50 V. A.C. £1.70 10mA £1.60 150 V. A.C. £1.70 20mA £1.60 300 V. A.C. £1.70 50mA £1.60 500 V. A.C. £1.70 100mA £1.60 8 Meter ImA £1.70		
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20mA \$1.60 300V. A.C. \$1.70 50mA \$1.60 500V. A.C. \$1.70 100mA \$1.60 8 Meter 1mA \$1.70		
50mA \$1.60 500 V. A.C. \$1.70 100mA \$1.60 8 Meter 1mA \$1.70		
100mA \$1.60 8 Meter 1mA \$1.70	20mA 21.60	300V. A.C. £1.70
100mA \$1.60 8 Meter 1mA \$1.70	50mA £1.60	500V. A.C. 21.70
100 meter. 22.10		
	200222 22 00	TO MCCOL BELLU

100mA &L'00	VO meter 32.10
Type MR.45P. 2	in. square fronts.
50μA £2.25	5 amp £1.70
50-0-50μA £2·10	10V. D.C £1.50
100µA £2·10	20V. D.C £1.50
100-0-100µA £1.87	50V. D.C £1.50
200μA £1.87	300V. D.C. £1.50
500μA £1.75	15V. A.C £1.80
500-0-500μA £1·70	300V. A.C. \$1.80
1mA £1.70	8 Meter 1mA £1.85
5mA £1.70	VU Meter £2.25
10mA \$1.70	1 amp. A.C. * \$1.70
50mA £1.70	5 amp. A.C.* £1.70
100mA £1.70	10 amp, A.C. *£1.70
500mA \$1.70	20 amp. A.C. *£1.70
1 amp £1.70	30 amp. A.C. *\$1.70

"SEW" BAKELITE PANEL METERS Type MR.65. 3in. square fronts. | 1 amp. ... 21 | 2 amp. ... 21



50µ 50¬ 100,

500

1m

1-0-

1001

Type Min.ou. of	m. square monts.
	1 amp £1.95
	2 amp £1.95
	5 amp \$1.95
Supplement of the last of the	15 amp 21.95
/ \	30 amp £1.95
	50 amp £1.95
	5V. D.C 21.95
	10V. D.C £1.95
	20V. D.C £1.95
	150V. D.C. £1.95
A £3.50	300V. D.C. £1.95
A £2.75	30V. A.C.* £1.95
0-50µA £2.35	50V. A.C.* \$1.95
uA \$2.35	150V. A.C.* £1.95
-0-100μA £2·25	300V. A.C.* £1.95
μA £2·20	500mA A.C.* £1.95
4 £1.95	1 amp. A.C. * £1.95
-1mA £1.95	5 amp. A.C. * £1.95
1 £1.95	10 amp. A.C. *£1.95
A £1.95	20 amp. A.C. *21.95
A £1.95	30 amp. A.C. *£1.95
mA £1.95	50 amp. A.C. *\$1.95
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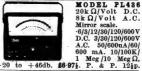
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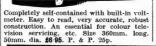




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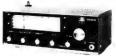


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everyday electronics

PROJECTS.. THEORY....

FOR ALL SEASONS

Our cover this month has quite an outdoor touch. Of course, you don't have to be an apiarist to sense that things are beginning to buzz in the outside world. Spring is now well advanced and thoughts are likely to be turning towards all kinds of pastimes and occupations for the coming summer months.

It is an appropriate time to point out that do-ityourself electronics has no closed season. Outdoor activities like gardening, touring, camping, sporting events, and so on, present many unique opportunities for putting electronics to effective use. So we advise, take stock now, anticipate your needs and start building to remedy any deficiencies in this respect.

GOOD COMPANION

The Constructors Companion given free with every copy of this month's Everyday Electronics is small and compact. It has been designed for your pocket, so that wherever you go you can have essential facts constantly at hand. Compiled with the beginner particularly in mind, this booklet will prove a valuable aide-memoire for the more experienced constructor as well.

Those still feeling their feet will be glad of the technical back-up they can instantly call upon when confronted with a choice of allegedly alternative or equivalent parts when shopping personally for components.

READY ACCESS

Our regular readers will already appreciate the amount of important and useful information they are accumulating, as the months go by. True, not everyone will have an immediate need for every project described. But a word of advice: do not discard back numbers. You never know when circumstances may arise that create a definite need which some previously described project would satisfy exactly.

This leads us on to another common problem: how to store numerous copies of a magazine so that ready access may be made at any time to one particular article. The only really satisfactory solution is to keep copies of the magazine in the binder specially designed to hold 12 issues of Everyday Electronics and which is now available.

fred Bennett

Our June issue will be published on Friday, May 19

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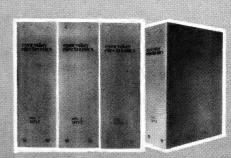
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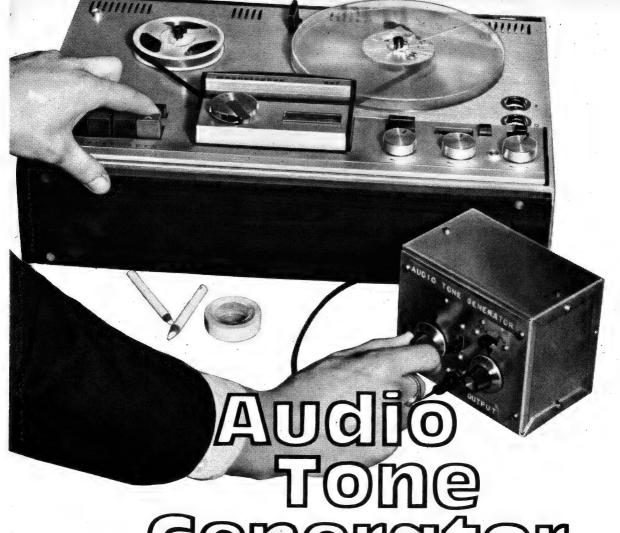
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Generator

BY FRED JUDD

This simple tone generator covers the audio range from 50 to 2,000 Hz and has been specifically designed for use with a tape recorder to make electronic music.

The multivibrator is one of the most commonly used electronic oscillator circuits and generates an almost square waveform. It can be made to cover a wide frequency range without the need for switching in different values of components and moreover will produce a high output signal level for a relatively small supply voltage. As a primary signal generator it has many uses as a test instrument in audio as well as electronic applications.

The generator described in this article is used as a tone source for the creation of electronic music and "science fiction" sound effects in conjunction with a tape recorder. The feature *Electronic Sounds and Music* on page 363 deals with the use of the tone generator in detail.

GENERATOR CIRCUIT

The circuit diagram is given in Fig. 1 and employs three *pnp* transistors, two of which form the multivibrator (TR1 and TR2), the remaining one, TR3, being used as a squaring amplifier.

The operating frequency and mark to space ratio (see Fig. 2) of the multivibrator are set by the time taken for C2 and C3 to charge up enough to switch transistors TR2 and TR1 respectively. This "charging time" is determined by the value of the capacitor and the value of the resistor through which it is charged.

Providing the time taken for each capacitor to charge is similar then the mark to space

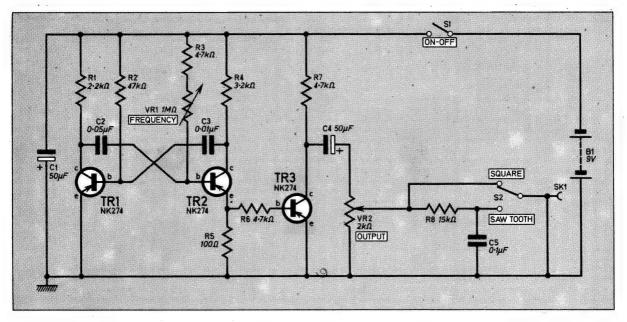


Fig. 1. Complete circuit diagram of the Audio Tone Generator.

ratio will be 1 to 1 or the mark and space will be of similar duration (Fig. 2). If we now change one of the controlling values—in this case VR1—both the frequency and mark to space ratio will be altered.

If we increase the value of VR1 the frequency will decrease as C2 will take longer to charge, and the mark to space ratio will alter for the same reason (see Fig. 3). Thus frequency control is achieved by VR2 and the total frequency range is approximately 50 to 2,000Hz.

The waveform has a mark to space ratio of 1 to 1 at approximately 1,500Hz at all lower frequencies the mark to space ratio increases becoming about 1 to 20 at the lowest frequency (Fig. 3).

The output from the multivibrator is taken from the emitter of TR2, through R6 to the base of TR3. Transistor TR3 is switched hard on and off by the output from TR2 and thus ensures a completely square output at its collector. The output level from TR3 is continuously variable from O to approximately 7 volts by VR2.

SAWTOOTH OUTPUT

The square wave output from TR3 can also be switched via S2 through an integrating network, C5 and R8, to provide an approximately sawtooth waveform (Fig. 4) of about 1 volt peak-to-peak maximum output, instead of the square-wave.

One of the major differences between a square wave and a sawtooth wave is the harmonic content and hence the tonal quality, when either are made audible via an amplifier and loudspeaker. The square wave contains only odd harmonics, in addition to its fundamental,

whereas a sawtooth wave consists of both odd and even harmonics plus the fundamental.

Audibly the square wave has a sound rather like that produced by a clarinet particularly in the region of middle C (261Hz approx.). The sawtooth wave has a sound rather more like a

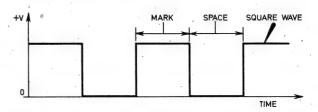


Fig. 2. A square wave with a 1 to 1 mark to space ratio.

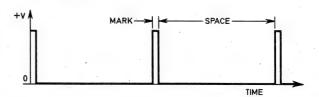


Fig. 3. A square wave with a 1 to 20 mark to space ratio.

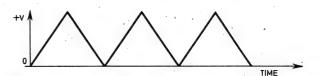


Fig. 4. A sawtooth waveform.

Audio Tone Generator

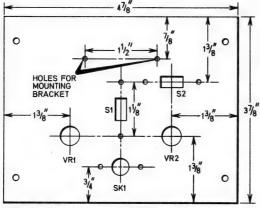


Fig. 5. Front panel drilling details.

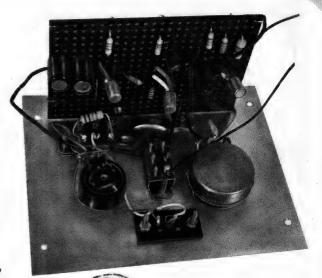
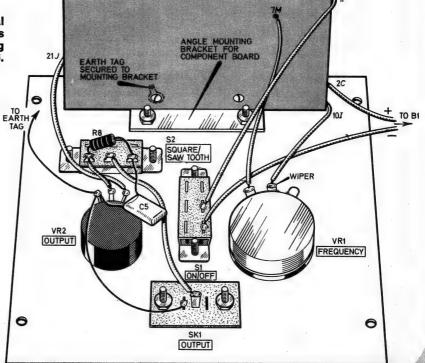




Fig. 7. Wiring of the final unit. The tinted area is the component mounting board as shown in Fig. 6.



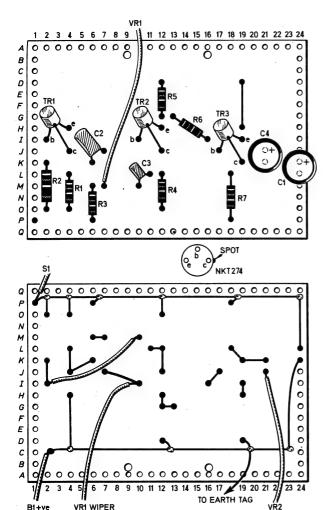


Fig. 6. Top and underside views of the component board. The transistor connections between the two diagrams are viewed from the underside.

flute. Both waveforms are used extensively in electronic organ voicing and for electronic music.

CONSTRUCTION

The prototype unit was housed in a box made from universal chassis parts. The pieces used assemble into a box measuring 5 by 4 by 3 inches. The sides and top and bottom can be assembled leaving one plate for the front panel and one for the rear. The plate used for the front panel is drilled as shown in Fig. 5 and is used to mount all the components.

If the layout and assembly of the generator is as shown there is just room in the case for a PP9 9 volt battery. Even if you spread the layout a little there should still be room for a slightly smaller 9 volt battery. The circuit board is 0.15 inch matrix plain perforated veroboard and is mounted on a 2 inch length of $^3{}_8$ by $^3{}_8$ inch aluminium angle.

Components....

Resistors

R1 2·2kΩ R2 $47k\Omega$ R3 $4 \cdot 7k\Omega$ R4 $3 \cdot 2k\Omega$ R₅ 100Ω R6 4 · 7kΩ $4 \cdot 7k\Omega$ R7 R8 15k Ω

SHOP TALK

All $\frac{1}{4}W$ $\pm 10\%$ carbon

Capacitors

C1 50μF elect. 12V C2 0·05μF C3 0·01μF C4 50μF elect. 12V

0.1 nF

Transistors

TR1 NKT 274 germanium pnp TR2 NKT 274 germanium pnp TR3 NKT 274 germanium pnp

Potentiometers

VR1 1M Ω log carbon VR2 2k Ω lin carbon

Switches

S1 S.P.S.T. slide S2 S.P.D.T. slide

Miscellaneous

SK1 Phono socket B1 PP9 9V battery

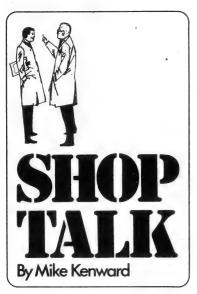
Control knobs (2 off) Eagle type F10, case 5 x 4 x 3in made from universal chassis panels or a similar size case, battery connector, aluminium angle 2 x $\frac{2}{3}$ x $\frac{3}{3}$ in. Veroboard 5 x 4 x 0 15in matrix plain perforated, earth tag, connecting wire, 4BA fixings.

Commence wiring of the component board by inserting all components except the transistors, and the wire link on the top of the board as shown in Fig. 6. Turn the board over and connect up the two supply lines along the two sides of the board using 18 or 22 s.w.g. tinned copper wire. Next connect up the remaining components using the component leads where possible and connect the flying leads.

Finally insert the transistors checking carefully the lead connections with the underside view shown in Fig. 6, and solder them to the other components using a heat shunt on each lead as it is soldered. After checking the circuit board mount the board on the aluminium angle bracket and mount this on the front panel together with the remaining components.

Wire up all the components to the circuit board as shown in Fig. 7 and check the completed unit carefully before connecting the battery and switching on.

Continued on page 386



This month we have one item which many readers will probably wish to construct but which is not given in the form of a constructional project. It is the simple passive mixer that is described and drawn up in the Making Electronic Sounds and Music feature.

Since this is really a bonus that will be useful to those following the article we have not given full constructional details or a components list. All the component values are given on the circuit diagram and the wiring diagram shows how they are put together. The three sockets can be any type suitable for use with your particular tape recorder—the types we have shown are phono sockets.

The complete unit can be mounted in any small case. No battery or power supply is necessary. We would like to emphasise that this is a simple passive mixer and will not be able to cope with all inputs.

A more advanced type of mixer may form the subject of a future article. However this simple mixer should be suitable for use with the *Audio Tone Generator* that is also described in this issue.

Audio Tone Generator

There should be very few buying problems for the Audio Tone Generator. As described above the sockets could be changed to any suitable type if your equipment does not use phono sockets or if you already

have other types. Once again the case for this project can be any type that is available in a suitable size.

Bee Counter

We find it difficult to comment on the availability of cedar wood —not after-shave—but apparently this wood must be used or the bees will not accept it!

As far as the remaining components for the *Bee Counter* go make sure that the resistors you buy are of adequate wattage. The lamp and holder should be of the miniature type so that they can be accommodated in the wooden base panel. Since the current drawn by this circuit is fairly large the section in the article concerning the battery should be noted.

There are a number of Post Office type counters available so make sure you get the right one —4·2 ohms coil resistance is the important thing.

Metal Locator

The Metal Locator is a project which we are sure will create great interest but please remember that this is a simple one-transistor design and cannot be expected to out-perform a £30 unit. The use of Perspex or Paxolin is recommended for the locator head as these materials are not affected by damp or water.

All remaining components for the locator should be readily available. The use of a subminiature switch is recommended since only a small hole then needs to be cut in the plastic beaker. Any $50\mu A$ moving coil meter could be used in the locator provided it will fit the beaker lid. The one specified is probably the cheapest.

Finally do not forget the operating licence and don't say we did not tell you!

New Products

Two products from one goahead firm have been introduced this month. Both in the audio field, possibly the most outstanding is the Unisound 505 as Radio and T.V. Components call their do-it-yourself £25 stereo system. This competitively priced unit comes as a complete kit and only needs two screwdrivers to put together. All the electronics are in module form and are supplied

with wiring looms that only need connecting up using a screwdriver supplied with the kit.

The large EMI speakers are housed in attractive cabinets again put together with only a screwdriver. It is said that anyone who can wire up a mains plug can put the system together in one evening. The system utilises modified Mullard Unilex modules, has an output of 3.7 watts continuous sine wave r.m.s. per channel; and frequency response of 40Hz to 20kHz at the 3dB down points. It would be very difficult to buy the individual components —including Garrard 2025TC deck, cartridge, plinth and cover and build a unit to match this one for £25, excluding the two speakers and cabinets.



The second unit from RT-VC is a £7 push button car radio kit, slightly more difficult to construct but any reader who has some experience of soldering should be able to build a working unit.

The kit is of good quality and uses the same push button tuning unit as radios costing three or four times the price. These features ensures good sensitivity and the pre-aligned i.f. (intermediate frequency) module and tuner avoid complicated alignment.

The kit is suitable for 12V positive or negative earth operation and readers may like to note that an after sales service—to repair any item not functioning correctly—is operated by RT-VC for all their kits; cost about £2 depending on the fault.





Simple experiments with a tape recorder

The term "electronic music" almost defies explanation because it is not the music that is electronic but the equipment and methods of creating it.

Its origin goes back many years, in fact to the invention of the thermionic valve and even as early as 1921 a "concert" of electronic music was performed in Paris by an Italian, Luigi Russolo, who used what was then called electrical sound generating and reproducing equipment.

Electronic music was difficult to perform directly from sound generators, etc., because composition required arranging the sounds in a given order and even changing the order, and sometimes the sounds, at a later time.

MODERN METHODS

Magnetic tape recording finally provided the ideal medium for composition. The sounds required could be recorded and rearranged afterwards by simply cutting out the pieces of tape containing them and splicing these together again in the order required. This technique paved the way for composers who, with both electronics and magnetic tape at their disposal, could produce new kinds of music with tonal qualities never before possible.

More recently of course the music synthesizer has taken over the task of tone generation, etc., and electronic music composers can now programme a synthesizer, couple it to a tape recorder and produce "instant" electronic music.

Nevertheless there is much that can be accomplished by the amateur with an ordinary domestic tape recorder, an audio tone generator (like the one described on page 358) and some splicing tape. The techniques are simple and you can get a good deal of fun out of experimental electronic music and "science fiction" sounds.

Your efforts need not be wasted either because you can enter them for the experimental music and sounds section of the annual British Tape Recording Contest (details later).

EQUIPMENT

An ordinary spool to spool tape recorder is the main requirement and if you have a stereo recorder with provision for recording independently on either track or you can get together with a friend and use two tape recorders, so much the better. A tape recorder with track-to-track or duoplay facility is also advantageous especially if it permits echo effects.

It is not possible to lay down procedures for specific makes and types of tape recorder but you will find that most of the techniques described can be applied.

Note that cassette or cartridge tape recorders are of limited use for creative recording of this nature which requires fairly extensive tape

cutting and splicing.

Most modern spool to spool tape recorders are designed for stereo operation employing half or quarter track on standard quarter inch wide tape. If the tape recorder has a track-totrack recording facility it will have separate recording and replay heads, thus allowing a recording on one track to be copied on to another together with other signals.

Some stereo recorders may only have a common record/replay head which will not normally allow track-to-track copying but may have a facility for making separate recordings on each of two tracks. Information concerning such facilities should be given in the tape recorder instruction book. If in doubt, you should contact your dealer or the manufacturer for such infor-

mation.

AUDIO TONE GENERATOR

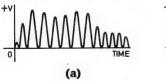
An audio tone generator is not absolutely essential but is most advantageous. The simple Audio Tone Generator described on page 358 is quite suitable as it covers a wide enough frequency range and will deliver a square-wave or a nearly sawtooth-wave output signal, thus providing two basic sounds.

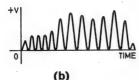
Sounds picked up by a microphone can also be used because these can be reshaped by tape cutting and splicing and by certain recording techniques. Magnetic tape will be required of course and for initial experimental work low

priced brands will suffice.

Some splicing tape and blank leader tape will also be required. Do not use ordinary plastic sticky tape, such as Sellotape, for splicing as

Fig. 1. (a) Original waveform of the recorded sound (b) The sound recorded and shown in (a) played in reverse.





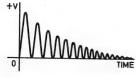


Fig. 2. Waveform of a sound which starts instantly and slowly dies away.

this may damage the tape and will not give a long lasting joint. Small kits of coloured leader and proper\splicing tape are readily available. A small tape splicer is also a very useful, though not essential, tool.

FIRST EXERCISES

It is important to know the extent to which your tape recorder can be used. If it has two or three speeds, as most of them do, record some musical sounds, whistling will do, at all three speeds and then play them back at one speed only, say the highest.

The sounds recorded at the lower speed(s) will be raised in pitch, by one or two octaves, depending on the speed. If the replay speed is double that of the recording speed the pitch is raised one octave and the sounds will occur faster but if the replay speed is half the recording speed, the pitch will be reduced by one octave and the sounds will occur slower. This is one of the most simple but most used techniques.

REVERSE REPLAY

Now, if your recorder is a stereo machine try turning the tape over (reverse the spools) and see if you can obtain replay on another track in reverse, i.e., the sounds will be going backwards. This technique is also commonly used for electronic music because it alters the nature of the sound completely by placing what was the beginning of the sound, i.e., its attack, at the end as illustrated in Fig. 1 in which (a) is the sound as recorded and (b) as played in reverse.

If you cannot play sounds in reverse try this exercise; connect a tone generator, or if this is not available record whistles through a micro-

Photograph showing the use of a tape splicer to join up a number of sounds.



Everyday Electronics, May 1972



Recording various sounds, using the microphone, to form a composition.

phone. Start with the recording level control at the maximum, set the tape running to record the sound but then almost simultaneously slowly turn the record level control to zero.

On replay you will have a sound that starts instantly and then slowly dies away as in Fig. 2. With a little practice you will be able to get various dying away or decay times depending on the speed at which the recording level control is turned off. Now try the reverse procedure; gradually increase the sound whilst recording and then quickly stop it.

TAPE CUTTING EXERCISES

Now try some tape cutting; first use the highest tape speed and record a few sounds of different pitch, i.e., from a tone generator, or whistles via the microphone, each one lasting two or three seconds.

Locate the beginning of each sound on the tape by carefully feeding the tape across the head and then cut the tape about two inches in front of the sound. Run off the remainder until you reach the beginning of the next sound; cut the tape here and splice to the end of the piece containing the first sound. Cut and join pieces of the remainder of the sounds.

On replay you will have a series of short sounds each rapidly following the other. Now try a similar exercise but this time insert pieces of blank leader tape between each sound.

MUSIQUE CONCRÊTE

Finally a variation of the two previous exercises. Record a few sounds each at a different tape speed. These should preferably be musical sounds, such as whistles or tones, or sounds produced by tapping a wine glass for example. Cut one or two pieces of each from the tape and assemble them at random with pieces of blank leader between groups. The pieces may be long or short.

Try replaying the assembled tape at different speeds and note the effect. You are well on the way to a form of composition known as "musique concrête" which is the creation of abstract forms of music out of real sounds. The same technique can, however, be used for abstract forms of electronic music in which the main sound source is an audio tone generator.

USING A TONE GENERATOR

The exercises outlined above demonstrate how almost any recorded sound can be altered by tape cutting and by recording and replay at different tape speeds. Electronic music does not normally include natural sounds recorded via a microphone and therefore the sound sources are electronic, i.e., from tone generators of one kind or another. The recording and tape cutting techniques, however, remain the same.

If you have a full range audio signal generator then tones can be recorded at the pitch required. The simple generator described on page 358 has a frequency range of approximately 50 to 2000Hz.

If frequencies outside the range of the generator are required it is simply a case of recording and replaying at different tape speeds for example; if a frequency of around 4000Hz is required, record the highest pitch of the generator (approximately 2000Hz) at a tape speed of 3_{4} in/sec (inches per second) and replay at 7_{2} in/sec.

If a very low pulsing sound is required at say 20 to 25Hz record a square-wave signal from the generator at its lowest pitch and then replay the recording at half the speed. Some experiment in this direction will soon reveal the tonal and pitch ranges that can be obtained simply by recording and replaying at various tape speeds.

Once this has been done, further experiment with the audio tone generator can be carried out in order to discover the type of sounds that can be produced. Start by recording a continuous note and while recording this vary the frequency and output controls on the generator, try this for both the square and sawtooth outputs (note that the output in the sawtooth position is much lower than in the square wave position).

Try cutting and reversing the sounds recorded to obtain various effects. You can also try making recordings at a distorted level by turning up the record level control, this will distort the original sound and produce yet another effect. Try switching from one output waveform to the other whilst recording—you can vary frequency and output at the same time—and also try switching the generator on and off while recording, again you can vary the output and frequency whilst turning on and off.

Edit the sounds produced by cutting and splicing and experiment fully with all possible

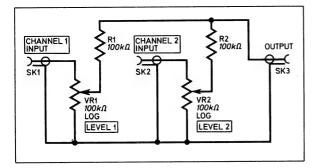


Fig. 3. Circuit diagram of a passive mixer that can be used for making electronic music.

effects. Once you have done this and feel fully conversant with the various effects that the generator is able to produce you can start to add one effect to another.

SIGNAL MIXING

Recording from track-to-track or using two separate tape recorders may necessitate mixing signals that are to be recorded and re-recorded i.e., signals from a recording already made to be mixed with signals from another source such as the tone generator.

Some recorders have built-in mixing facilities whilst others may permit a form of mixing by using the track-to-track recording facility or by superimposing one sound on another previously recorded. Again the tape recorder instruction book will provide information of this nature.

However, it is possible to build a very simple mixing circuit as shown in Fig. 3; Fig. 4 shows the construction. This is known as a passive mixing network, but will alow two signal sources to be mixed at different levels and coupled to a common input on a tape recorder (Fig. 5).

TAPE LOOPS

Another interesting technique widely used for electronic music is the tape loop. This is the use of a small endless loop of tape containing recordings which are played continuously to produce repeating rhythm patterns.

Record a few natural sounds, or low pitched tones from an audio generator, of quite short duration, one immediately after the other. Cut a piece of the tape containing the sounds, about 18 inches long, and splice the ends together so as to form a complete loop. Place the loop in the recorder so that it runs past the tape heads when the machine is set to replay. You can hold the loop under tension by one of the methods shown in the photographs. Try running the loop at different speeds and, if possible, reverse the direction.

Record some percussion sounds, e.g., sounds produced by knocking together empty boxes, etc. Cut out pieces and make up a loop consisting of the various sounds and blank leader tape.

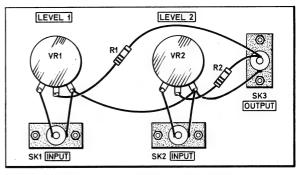


Fig. 4. Constructional details of the circuit shown in Fig. 3. Shop Talk refers to this figure.

For the first attempt use only two or three sounds and two or three pieces of leader.

You can make up an almost endless variety of fascinating rhythm patterns by this method and if you use two tape recorders the rhythm loop can be copied from one to the other whilst other sounds are added.

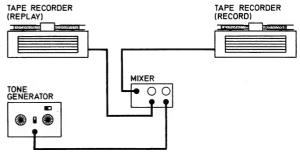
MULTIPLE RECORDING

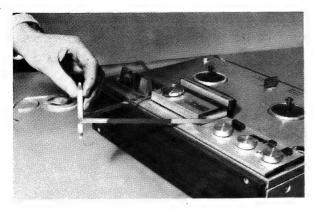
If you have a tape recorder with a track-totrack recording facility the scope is much wider as sounds may be recorded on one track and then re-recorded on to another track whilst adding more sounds. If your tape recorder can produce the echo effect this too can be used in various ways to produce those echoing science fiction sounds. Try allowing the echo to build up into a crashing roar and see if you can play it in reverse.

Now that you have discovered the variety of sounds and rhythms available using the facilities you have it is up to you to put these together to form an interesting "musical" passage. It may take some time before you achieve the required effect.

By combining even a few of the techniques outlined the number of permutations possible are fantastic. Instructions on composition cannot be given because no rules exist. Your ideas must come solely from imagination and experiment.

Fig. 5. Using the passive mixer to combine two signals for recording purposes.









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The three photographs on the left illustrate various methods of using a tape loop. The top photograph shows a reversed loop held under tension by passing it around a pencil; this is only suitable for short periods.

The centre photograph shows a reversed loop held under tension by a small spool hanging over a table edge; this is only suitable for fairly large

loops.

The lower photograph shows a system that can be used for any size loops by routing the tape around suitable objects—batteries are shown. This photograph also shows a cardboard tape holder used to keep recorded sections of tape in the order required.

COMPETITION

Finally, why not try an entry for the "technical experiment class" of the annual British Tape Recording Contest. It is open to anyone and the closing date for the 1972 contest is not until June 30. The Technical Experiment class allows for tapes of up to 4 minutes duration and includes; sound composition, electronic music, musique concrête, multi-track music and experimental sound recordings. The prizes are worthwhile and you can get an entry form free by writing to The Secretary, British Amateur Tape Recording Contest, 33 Fairlawnes, Maldon Road, Wallington, Surrey, and enclosing a stamped addressed envelope. You may also be interested to know that the special "Tape of the Year" award for 1971 was for an experimental class entry.

Every tape entered is carefully assessed by the expert judges and their comments are passed to the contestant concerned when the tape is returned. Thus you will know how to make an even better tape next time.

PLEASE TAKE NOTE

The approximate cost of components given in the Simple Calculator article last month was incorrectly shown as £1.20. This should have been £2.20.

The probe flying lead in the Signal Injector article (March issue) should be soldered to Y3 not Y2 as stated in the text.

The Normatest 2,000 multi-range test meter mentioned in Shop Talk last month is available from: Croydon Precision Instrument Company, Hampton Road, Croydon, CR9 2RU.



THEY MADE THEIR MARK

NO1 Introduction By J. E. Gregory

ELECTRONICS is an internationally uniform world of symbols. Look at any advertisement or study the simplest circuit diagram in EVERYDAY ELECTRONICS and you will be confronted with strange symbols of every shape. Magical signs used to signify basic units of physical quantity; Table 1 lists some of them.

Although electronics is regarded as a modern science and hobby many of these units are named after pioneers, scattered throughout the world, whose accumulated research spans hundreds of years.

This series sets out to explain the symbol, and perhaps more important something of the man who gave his name to it. But let's begin our potted history of electronics at the beginning.

THE GREEKS HAD A

Take the word electronics itself, for that we must go back in time to ancient Greece. To the ladies of Greece passing time by decorating their spinning wheels with amber, found on shores in the far north. They observed that the amber when contacting the threads would draw the threads to itself as they separated from the wool, and then push them away in a frictional force. The

Greek word for amber was elecktron, from the verb elkein to attract. Although this phenomenon was observed and noted by several of the great Greek philosophers we have to jump two thousand years to the early 1600's and to the reign of Good Queen Bess, who was persuaded by her physician William Gilbert, to attend a demonstration of a frictional electric machine based upon the power of amber to attract. This power he called electricity.

electricity.

It was soon realised that the crackling and sparking of Gilbert's electric machine were the same phenomena on a minute scale, as thunder and lightning, but how to prove it?

THE KITE FLYER

One of the first to try was the fifteenth child of an English immigrant; born in Boston Massachusetts in the year 1706, this was the well known American statesman and philosopher Benjamin Franklin (see illustration above).

His historic but dangerous

experiment trying to capture electricity from the sky occurred during a thunderstorm in the summer of 1752, when accompanied by his small son, he flew a kite with an iron door key. During the storm, he saw that sparks sprang from the key to his wrist, what he didn't realise of course was that if the lightning had actually struck the kite he would have been killed.

The study of natural phenomena had to take second place to his other activities, but he came to the conclusion that thunderstorms were simply the levelling of opposed electrical potentials, between one cloud and another or between a cloud and earth.

It was Franklin who introduced the positive and negative signs for electric charges, realising there are two kinds which neutralise each other.

Next month we move from America to 18th Century Italy and a scientist, Alessandro Volta, after whom the Volt, the measurement of electrical potential is named.

Photograph: Science Museum, London.

Table I FUNDAMENTAL UNITS

unit symbol	name of unit	physical quantity	
* A	Ampere	Electric Current	- 7
V	Volt	Electric Potential	. 198
F	Farad	Electric Capacitance	
Ω	Ohm	Electric Resistance	
W	Watt	Power	
Hz	Hertz	Frequency	
H	Henry	Inductance	The same

These basic units are often inconveniently large or small and the units are prefixed with the following symbols:

	11/2	9		pico				00 milli	on	
	9 1	n		nano				nillion		
	1			micro)		milli	on		
		m		milli kilo		÷ I × I			age of the second	All Street
1310	1200	м	Alexander	mega			milli	on		
		G		giga				nillion	100	

Hence 5kV = 5,000 Volts; or 5mV = 0.005 Volt





BEGINNE FOR

By Mike Hughes M.A.



HIS year sees the twentieth birthday of the component most responsible for bringing electronics within the scope of do-it-yourself enthusiasts; it has greatly simplified design and construction and has also brought about terrific reductions in costs. It is the "transistor".

As a replacement for the valve, it allows us to use low voltages and removes the arduous task of having to assemble valve bases and massive transformers on tank like chassis. Connections to a transistor are few and the basic way it operates in a circuit is quite easy to understand.

PNP-NPN

The transistor is a member of the semiconductor family and is basically a sandwich of different types of either silicon or germanium. The "filling" of the sandwich can either be p- or n-type material: we can clad a p- type filling with ntype material giving what we call an npn transistor. Alternatively a pnp device is made by filling a p-type material with an n-type.

One encounters both types in practice but nowadays npn devices made from silicon predominate, the reason being that they are easier to make and hence cheaper!

Fig. 1(a) shows a diagramatic cross-section of both types of transistor, pnp and npn. One end is heavily doped and is called the "emitter"; the other end is lightly doped and called the "collector".

The filling material is very thin in practice (usually one or two microns; 1 micron is a millionth of a metre) and is called the "base". In its simplest form you can think of an npn device as two diodes connected together by their anodes (back-to-back), and facing each other in a pnp device, Fig. 1(b).

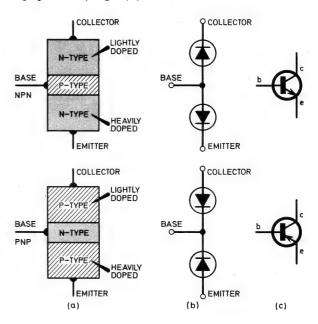
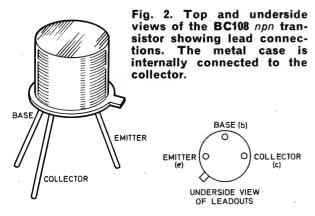


Fig. 1. (a) Schematic diagram of the internal make-up (b) equivalent representation and (c) circuit symbol for (top) npn transistor and (bottom) pnp transistor.

BASE CONNECTIONS

All the transistors you will come across have connections brought out from the emitter, base and collector. A very common silicon *npn* device is the BC108 and we shall be referring to this frequently in this series.

Fig. 2 shows what it looks like. If you have one handy see if you can identify which lead is which.



The emitter is the one closest to the spigot on the side of the can, the collector is diametrically opposite, and the base is between the two but set off to one side. This is a metal can transistor and the can is electrically "live"—in actual fact it is connected to the collector as well as the lead out wire.

Different types of transistor may have different shaped cans and some are in plastic encapsulations. Always make sure you know which lead is which before you start using a transistor.

Most constructional projects in EVERYDAY ELECTRONICS give you lead designations for the transistors specified, but if you want to experiment with alternative types make sure you know the correct base lead connections.

SIMPLE TEST

Use the BC108 npn transistor to identify the effect of the two diodes connected back-to-back. First of all make an ohmmeter on the Demo Deck. Use a 4.5V battery (not 9V) in series with a 2.2 kilohm resistor and VR2 (5 kilohm). Complete the circuit and set VR2 to give zero ohms at full scale deflection and then connect the leads of your ohmmeter between the base and emitter connections of the transistor—to do this it is best to solder the transistor on to three adjacent pins of the Demo Deck and use crocodile clips on the leads from the meter.

If you connect the meter so that the lead coming directly from the negative terminal of the battery goes to the emitter, the meter needle will move to almost full scale showing there is little resistance in the transistor. Now reverse the leads so that the base is more negative than the emitter—you should see that no current

flows (indicated by meter needle not moving). Thus the base-emitter junction is a diode and follows the same rule that we saw last month.

Now leave the lead on the base and transfer the one from the emitter to the collector—again no current flows but reverse the leads and current flows between the base and collector.

If you connect the leads between the collector and the emitter no current should flow whichever way you have them because in both connections, the current would have to pass through a reverse biased diode.

This simple experiment can be used as a rough and ready test to check if a transistor is likely to be in working order, and provided you remember the rule "make p stand for positive for current to flow" you can use it to identify npn and pnp transistors.

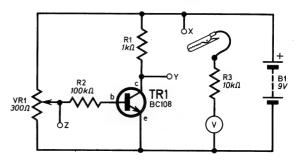
REVERSE VOLTAGE LIMITS

Like all diodes, the junctions of a transistor have reverse voltage limits. These are usually specified with abbreviations. For the BC108 the reverse emitter-base voltage ($V_{\rm ebo}$) is 5V—i.e. you must never make the base more than 5 volts negative with respect to the emitter (this is why we had to use $4\cdot5V$ for our ohmmeter instead of the 9V we have been used to). Likewise the reverse base/collector voltage ($V_{\rm cbo}$) is 30V. You might expect the reverse voltage between the emitter and the collector to be equal to the highest of the other two but this is not the case—it is lower—for the BC108 $V_{\rm ceo}$ is 20V.

The "O" in the suffixes of the reverse voltage characteristics indicates that the third terminal is "open circuit" i.e. not connected.

HOW THE TRANSISTOR WORKS

Let's see what a transistor actually does by using the circuit of Fig. 3(a). Now that we are using the transistor in a real circuit it is important to note the polarity of the supply voltage—for an *npn* transistor the collector must always be kept more positive than the emitter (the converse applies to *pnp* devices). We are going to make the transistor work like a tap and control the amount of current flowing through R1. You can see this happening if you follow the details through on the Demo Deck.



VR1 is a 300 ohm potentiometer working as a potential divider giving us a variable supply at its wiper.

Wire up the circuit of Fig. 3(a) on the Demo Deck as shown in Fig. 3(b), but do not connect R2 to the base of the transistor just yet.

Resistor R3 and the 1mA meter makes a 10V range voltmeter in the usual way. Connect the negative lead to the emitter of the transistor. All voltages we measure will be relative to that of the emitter.

First measure the power rail at point X—it should, of course, be +9V; now measure the potential at the collector of the transistor (point Y) it should be $+8\cdot 2V$. This is what is expected because no current can flow through the back-to-back diodes of the transistor, but the meter will draw some! If you had a high sensitivity meter (say 20 kilohm per volt) this current would be negligible and you would see +9V at both points, X and Y.

Now set VR1 so that the potential on its wiper is zero (with respect to the emitter) and connect R2 to the base of the transistor. VR1 potential is measured by attaching the crocodile clip from the meter to point Z. Again measure the potential at the collector—it should not have changed.

We shall now see what happens if we increase the potential at the wiper of VR1. Do this in 0.5V increments (use crocodile clip at point Z) and for each setting measure the collector potential. You should see that once the potential of the wiper exceeds 600mV, the potential at the collector falls, and continues to fall towards zero as the controlling voltage is increased. Once the collector potential reaches almost zero no more

control can be effected. We say that the transistor is now fully conducting between collector and emitter. This state is called "saturation."

Record your results and plot a graph of collector voltage versus voltage at the wiper. A graph should be obtained similar to that of Fig. 4.

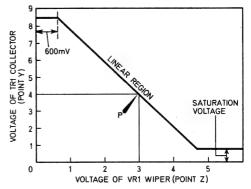


Fig. 4. The graph obtained by plotting the recorded results of experiment using circuit of Fig. 3(a), i.e. voltage at point Y versus voltage at point Z.

Control of the collector/emitter current is brought about by passing a current through the forward biased base/emitter junction. The more current we pass into the base in this way, the more current we can control between the collector and the emitter. The controlling current is called "base current," (I_b) and the controlled current "collector current," (I_c) .

Base current is set by the potential difference between the wiper of VR1 and the emitter of the transistor, acting through the resistance R2

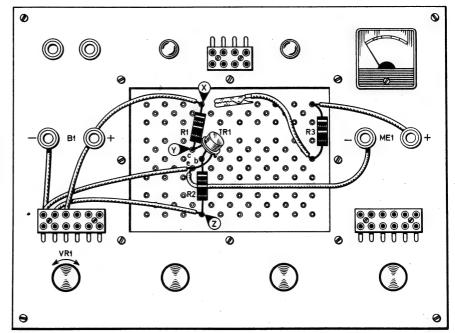


Fig. 3(a) (left). The circuit diagram used for investigating some of the properties of a BC108 transistor.

Fig. 3(b) right. The circuit of Fig. 3(a) wired up on the Demo Deck.

and any internal resistance between base and emitter. The latter is small and can be neglected at this stage. We must, remember, however, that the base must be made at least 600mV positive with respect to the emitter before any current can flow (this is the usual forward voltage drop for any silicon junction).

We can thus calculate the current flowing into the base by measuring the potential at the wiper of VR1, subtracting the base emitter forward voltage drop (600mV) and dividing by the value of R2.

GAIN

If you do this for your experiment you will find that the base current ranges from 0 to 0.084mA. The range of collector current we are controlling was from 0 to 9mA. It can be seen that the transistor enables us to use a very small current to control a larger one. We call this effect "current amplification." The factor that governs the ratio between I_b and I_c is called "gain" and although it increases with Ic it is pretty well constant for any given transistor. It can, however, vary widely between different types of transistor and even between devices having the same type number! Provided you take a combination of base and collector currents within the controllable region (this is called "linear region") you can calculate the gain of the transistor you are using.

It would be best to increase the potential at VR1 until the collector potential is approximately 4V. This reduces the shunting effect of our voltmeter.

Use the precise values of voltage measured to calculate the current through R2 and R1 then use the ratio of these values to calculate the gain.

gain = collector current \div base current = $I_c \div I_b$

For the BC108 transistor it should be approximately 200, but as we have said, will vary from device to device.

Example To calculate the gain from your plotted curve (similar to the one of Fig. 4) select a convenient point on the linear region such as point *P* of Fig. 4.

The base current, I_b is given by the voltage difference between the base and emitter divided by the base resistor.

i.e.
$$\frac{3-0.6}{100.000} = 0.024 \text{mA}$$

Now the voltage drop across the collector resistor R1 is (9-4)V=5V. Therefore, collector current I_c is $(5\div1000)=5$ mA.

Substituting these values for I_c and I_b in equation (1) gives the gain = $(5 \div 0 \cdot 024) = 208$.

There are various ways of describing current gain for a transistor so we shall define that measured above a little more precisely—it was the d.c. current gain. This is sometimes abbre-

viated to the designations β (beta) or $h_{\rm FE}$. The latter is rather a strange type of designation but is one of a range of what are called "h" parameters—we need not worry ourselves about these in this series except for the term $h_{\rm FE}$ which is usually used in manufacturer's data sheets. Do not confuse $h_{\rm FE}$ with $h_{\rm fe}$, the latter is called the small signal current gain and we shall not be dealing with this until later.

The gain equation above can be rewritten: $I_c = h_{FE} \times I_b$

Remember that the experiment we have just done has been using a silicon npn device. We could have used one made from germanium having npn structure and obtained a similar effect—except that the base/emitter forward voltage drop would have been only about 200mV and $h_{\rm FE}$, in general, would have been lower.

We could also have used a silicon or germanium pnp device but would have had to reverse the battery connections so that the collector was negative with respect to the emitter. The same rules would have applied and we could have still calculated a value for $h_{\rm FE}$.

If you are a little confused by the difference between npn and pnp devices do not worry too much as this stage—most of the early experiments in Teach-In will use npn devices and when you have got used to these you will find it quite straightforward to switch over to pnp devices when necessary. The most important thing to remember is the polarity of battery voltage when using one type or the other. An aid to remembering what the polarity ought to be is to bear in mind the direction of conventional current flow;

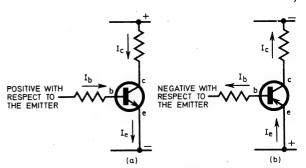


Fig. 5. Circuits showing major current flow directions for (a) npn and (b) pnp transistor. I_b —base current, I_c —collector current, I_e —emitter current.

the arrow on the emitter of the symbol points in the direction of current flow, i.e. it points away from positive and towards negative. See Fig. 5.

Whether using npn or pnp devices an aid to remembering how to turn collector/emitter current "on", is to make the potential at the end of the resistor connected to the base tend towards the same polarity voltage as applied to the collector; the more you move towards this voltage, the more I_b increases, and I_c will increase in direct proportion.

When the potential feeding the base rises towards the supply voltage the voltage at the collector falls towards the emitter voltage. This is called "inversion."

In Fig. 3 R1 is called the "collector load." The limit of I_c control is set by the value of this resistor; if it has a high value then it does not matter how much base current you apply, you cannot control more collector current than that given by the collector supply voltage divided by the value of collector load. On the other hand, if the load is too low you might find yourself trying to force more collector current than the construction of the transistor can handle. Thus one of the specifications of a transistor is the maximum collector current it can handle without "blowing". This is called I_{cmax} and for the BC108 is 100 mA.

A final parameter we must deal with is the power rating of a transistor. As current is passing through it a certain amount of heat is dissipated. We already know that too much heat can spoil the properties of a semiconductor so it must be limited. The limit is set by the maximum power dissipation parameter, $P_{\rm cmax}$. It is easy to calculate what the power dissipation is likely to be; it is the dissipation you would get if you replaced the transistor in the circuit with a resistor having the same ohmic value as the collector load.

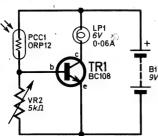
Table 1 gives you some typical values of parameters for some common transistors of varying types, powers and polarities.

Table 1: THE MORE IMPORTANT CHARAC-TERISTICS OF SOME COMMON TRANSISTORS

Type P	olari	ty P _{c max}	V_{cbo}	V_{ceo}	V_{ebo}	Ic max	h _{FE}
BCI08	nþn	300mW	30V	20V	5V	100mA	240
2N2926	nþn	200mW	18V	18V	5V	I00mA	150
BFY51	nþn	800mW	60V	60V	6V	IA	70
BFX13	pnp	300mW	-20V	-15V	5V	I00mA	120
2N3702	pnp	360mW	-40V	25V	-5V	200mA	60
ACI26	ÞnÞ	500mW	-32V	-32V	-10V	I00mA	100
OC72	þnþ	125mW	-16V	-16V	-3V	125mA	50
OC26 -	ÞnÞ	12W	-16V	-16V	-107	3.5A	50
OC36	þnþ	30W	-80V	-32V	-40V	10A	70

Fig. 6(a) (below). The circuit diagram of the "Electronic Candle" which illustrates positive feedback.

Fig. 6(b) (right). The circuit of Fig. 6(a) wired up on the Demo Deck. Ensure that PCC1 is close to LP1.



ELECTRONIC CANDLE EXPERIMENT

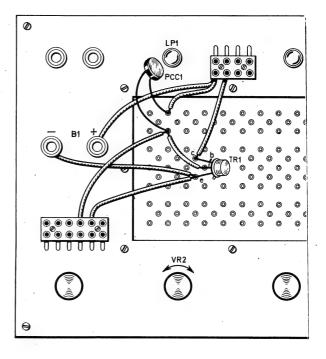
We shall now make a simple working circuit using the circuit diagram of Fig. 6(a). This is wired up on the Demo Deck as shown in Fig. 6(b). Connect the ORP12 (light dependent resistor) very close to the LP1 on the Demo Deck as shown below. Set VR2 to zero ohms. The potential at the base of TR1 will be zero, therefore no current will flow between collector and emitter. Now, in a reasonably lit room, increase the value of VR2. At a certain point the potential at the base will reach 0.6V (set by the potential dividing effect of PCC1 and VR2) and the transistor will start to conduct (the bulb will glow dimly).

Continue to increase the resistance of VR2; the current flowing through PCC1 will now pass into the base/emitter circuit of the transistor in preference to the higher resistance path through VR2. This base current will cause TR1 to pass more collector current until the bulb is fully illuminated.

When you reach this point (the minimum value of VR2 that will give full illumination) try casting a shadow over PCC1, the lamp will go dim and ultimately go out altogether as I_b reduces due to the resistance of PCC1. We did a similar sort of thing in Teach-In Part 4.

The difference is that we now have a circuit that is much more sensitive to small changes in light level which is brought about by the transistor amplifying the current from the photo resistive cell.

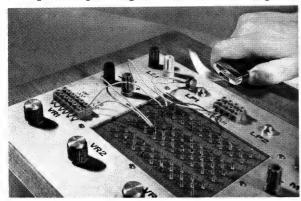
If you place the cell very close to the bulb in a dimly lit room you can set the value of VR2 so that the ambient lighting does not turn the transistor on, but the light from the bulb will.



Break the light path between the bulb and the cell and the bulb goes out and stays out. Now use a match or lighter to provide a stimulus of light. Bring it close to the bulb/cell assembly and the bulb lights up; you can now remove the match and the bulb will stay on because its own light output is holding the transistor on. This is called "positive feedback" and in this circuit will provide an amusing party trick—especially if assembled to look like a candle.

A bit of practice at "snuffing" the candle with the fingers (actually you are breaking the light path between the bulb and the cell) will make the effect even more astounding.

Photograph of the Demo Deck set up for the Electronic Candle Experiment showing the lamp being "lit" by the light emitted from the lighter.



TEACH-IN PART 6—ERRATA

Fig. 4(b) last month shows a lead connected wrongly. The lead from the junction of R3 and the negative meter terminal should go to the negative end of VR1 (not the wiper as shown) i.e. the one connected to the battery negative.



Next month: Multivibrators. The components needed for next month in addition to those already acquired are: resistors 22 kilohm (2 off), 100 ohm (1 off); capacitors $0 \cdot 1 \mu F$ polyester (2 off), $500 \mu F$ elect. 12V (1 off); transistors BC108 (1 off); diodes OA91 (1 off).

Ruminations By Sensor

Not so Clever

The coal miners' strike has shown how dependent we are, in this age of high technology, on the efforts of men who work in damp, dirty and often dangerous conditions.

I find it difficult to comprehend that on one hand the semiconductor industry owes its existence to the ability to obtain and to process materials with an impurity content of less than ten parts in a thousand million, and to operate with tolerances down to one millionth of a metre, while on the other hand men have still to dig fossil trees out of the earth (albeit with mechanical assistance) so that these fossilised remains can be burnt to boil water in order to raise steam

and to generate electricity! Without coal and electricity there would be no semiconductor industry; truly our idol has feet of clay!

Let There be Light

Have you heard about the old lady who telephoned the C.E.G.B. to complain that, during the power cut, the buses were passing her house with all their lights on? She also said that she could manage to get along quite well without the electricity, except for the little light in the hall, and could they please leave that one switched on.

Many people must have been irritated, in the early days of the strike, to see street lights blazing all day and switched off at night, due to their electric clock switch mechanisms getting umpteen hours behind. To the electronics man the answer to this problem is so simple—a light operated switch, either using discrete components or in integrated form.

A recently introduced inte-

grated circuit provides the necessary photo cell, level sensor and time delay all on one tiny chip of silicon and complete with lens. It could operate a relay or, better still, work into a switching transistor controlling the street lamp directly.

Some years ago, I was shown around a large generating station, where, tucked away in a dusty corner there was a cast iron box about the size of a domestic cooker. This apparatus was installed at the station about twenty five years ago and its purpose was to switch on all the electric street lamps in the town.

When switched on it produced a ripple which was superimposed on the mains. Sections of street lighting were grouped together under the control of master switches, spread throughout the town, which were operated by switching on the ripple equipment. The system had been in use but for some reason, unknown to my guide, had been discontinued. It would have been a blessing during February 1972.

Everyday Electronics, May 1972



Multimeter

Probably the most useful of all test equipment is a multimeter and next month we show you how to build a fairly simple one that will meet the needs of most constructors.

Light to Sound Converter

A project for those who like to experiment. This unit produces an audio tone, the frequency of which is dependent on the light level sensed by a photocell.

Also...

A new feature for all

beginners and constructors... GUIDE TO CIRCUIT SYMBOLS

A new feature for all beginners and constructors; Guide To Circuit Symbols. We explain the symbols and show you what the components look like. Starting next month.

On sale Friday, May 19



M ODERN research calls for accurate measurement and comparisons, and with this in mind this device was designed to help the beekeeper assess the performance of his beehives more definitely, and to compare the different strains of bees under the same working conditions and so help to breed a strain which will produce the most honey under all the difficulties encountered in our changing climate without the rather nasty habit of the English bee, of attacking the bee-keeper as soon as he appears anywhere near the hive.



The Bee Counter is an instrument which records the number of bees entering the hive, and used in conjunction with other devices such as a wind speed indicator, a wind direction indicator, an air temperature thermometer, a maximum/minimum thermometer, a rain gauge and a sunshine recorder, then some degree of assessment can be made, and some basis established for the bee-breeder to work upon his main goal—lots of honey from a reasonably good tempered, busier bee.

The Bee Counter makes use of the fact that bees are highly organised in their habits, and utilises the bees sense of sight and smell. These bee "characteristics" are used in the design of the cabinet housing all the circuitry which is described later in full detail.

THE CIRCUIT

The complete circuit diagram of the counter is shown in Fig. 1 and is basically an amplifier which works as follows.

The lamp LPI, which is always "alight" when

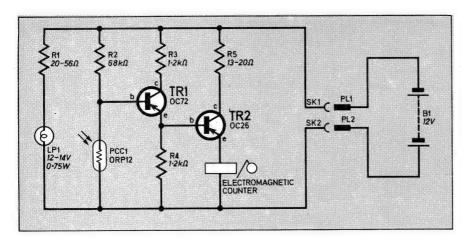


Fig. 1. The complete circuit circuit diagram of the Bee Counter.

the unit is switched on, illuminates the light dependent resistor, PCC1, and causes its resistance to be at a low value, about 100 ohms.

The l.d.r. and R2 form a potential divider circuit and under "illuminated conditions" of the l.d.r., a positive voltage with respect to the emitter, is applied to the base of TR1 causing it to be in a conducting state.

With TR1 conducting, a negative voltage is applied to TR2 base with respect to the emitter and consequently TR2 is "off" (not conducting).

When the light path between LP1 and PCC1 is broken, the resistance of PCC1 increases considerably (to about 100 kilohm for complete "blackout"). This causes the potential at TR1 base to go negative and turns it "off". This state of TR1 causes the voltage applied to the base of TR2 to go more positive and causes it to switch "on" i.e. conduct—current flows through TR2.

When current flows through the emitter leg of TR2 containing the relay coil in the counter, the relay is energised.

When the light to PCC1 is restored, TR2 switches "off" and the counter is de-energised and springs back to its off position and in doing so mechanically adds "one" to the counter readout.

The arrangement of LP1 and PCC1 in the case is so devised that the bee, on entering the hive, breaks the light path between these devices and its entry is thus recorded.

The 13-20 ohm 3 watt resistor, R5, in the collector circuit of the power transistor, TR2, is to prevent damage to the counter or the transistor if the entrance passage to the hive should become blocked, as once happened in the prototype when a drone got stuck in the narrow part.

A heavy duty battery is required to operate the Bee Counter since current drain is substantial —250 mA when TR2 is "off" and 400 mA when TR2 is "on" at 12V. A car battery is therefore recommended to supply the power. The cost of this battery is not included in approximate cost.

The voltage is fairly critical as it must be sufficient to operate the counter, but not high

Components....

Resistors

R1 20-560 3 watt

R2 68kΩ

R3 1·2kΩ

R4 $1 \cdot 2k\Omega$

R5 13–20 Ω 3 watt

All $\frac{1}{2}$ watt carbon $\stackrel{...}{\underline{...}}$ 10% unless otherwise stated

Transistors

TR1 OC72 (or similar) germanium pnp TR2 OC26 germanium pnp

Light Dependent Resistor

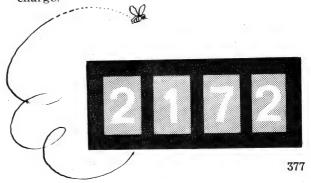
PCCI ORP12

Micellaneous

LP1 12-14V 0·75W bulb and holder PL1, PL2 Wander plugs, 1 red 1 black (2 off) SK1, SK2 Sockets to suit plugs PL1, PL2 B1 12V battery—heavy duty rechargeable type (Not accounted for in cost box.) Counter: Post Office type 14C 4·2Ω 4 figure readout. Cedar wood, Perspex and adhesive, Paxolin, wood screws, 4 B.A. nut and bolt, wood glue.

enough to cause overheating of TR2 or the counter coil in the event of the passage being blocked for long.

If the apparatus is disconnected every night the battery will last at least a week on one charge.



Everyday Electronics, May 1972

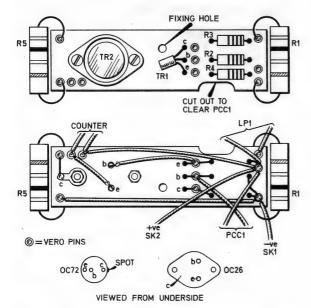


Fig. 2. The layout of the components on both sides of the Paxolin board. Veropins are used for attachment.

Variations in performance can be dealt with in several ways. The lamp should be bright enough to turn off the amplifier, but not any brighter than necessary. This is best adjusted by altering the series resistor R1, which may be increased to as high as 56 ohms.

Also, the size of the light hole can be varied, or a part of the l.d.r. painted over so that it has less area exposed, until the instrument is sufficiently sensitive, but positive in its action.

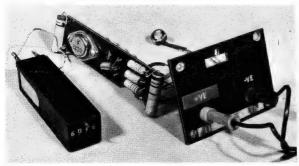
THE COUNTER

The electromagnetic counter used is a Post Office type. It has a four digit readout and can thus count up to 9,999. The maximum count rate is ten per second.

COMPONENT WIRING

Most of the components of Fig. 1 are mounted on a piece of Paxolin size $4^{1}_{2} \times 1^{1}_{4}$ inches with a cut-out as shown along one side to accommodate the light dependent resistor, PCC1.

Both sides of the board containing the components are shown in Fig. 2.



Veropins are used for mounting the components in position and small holes should be drilled where indicated to accommodate these pins.

Three more small holes of the same size should be drilled to take the leads of TR1 as shown.

Drill the component board fixing hole and the four holes for transistor TR2; (see reverse side of component board Fig. 2); ¹₈in. diameter holes will do for all five holes.

Begin assembly by pushing in all the Veropins and then attach TR2 to the board using two small nuts and bolts.

The connection to the collector of TR2 is via its casing, so a solder tag should be attached to one of the securing bolts to enable this connection.

Attach and solder all the components, link wires and flying leads as detailed in Fig. 2 making sure a heat shunt is used when soldering in TR1, which incidently should be the last component connected.

The l.d.r. should be attached to the board via 6in. long flexible leads.

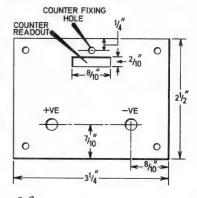
The flying leads to the counter should be about 4in. long.

The two wander sockets used for battery connection to the counter, are attached to the end of the case which is made from a piece of Paxolin, dimensions are given in Fig. 3.

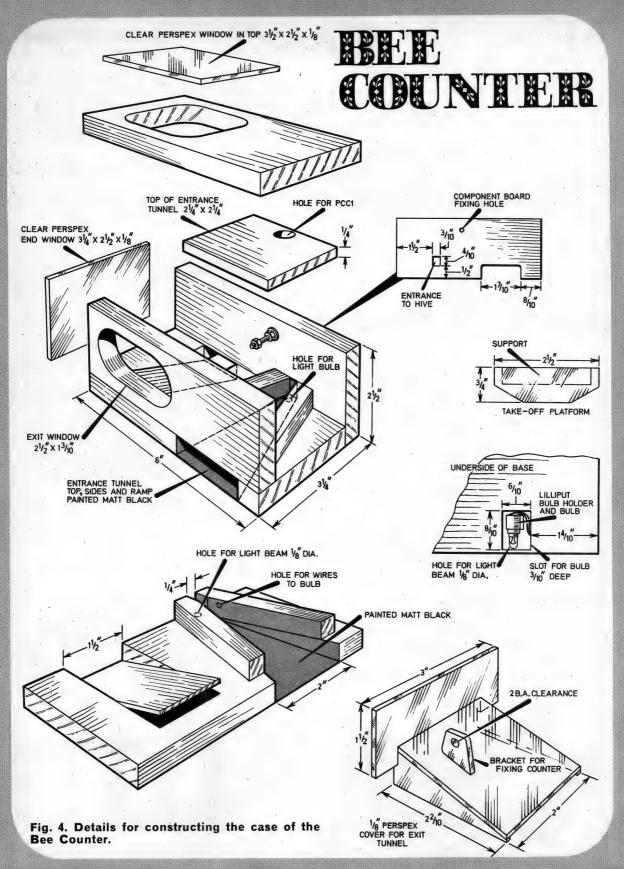
The connection wires from the wander sockets to the component board should be about 4in. long.

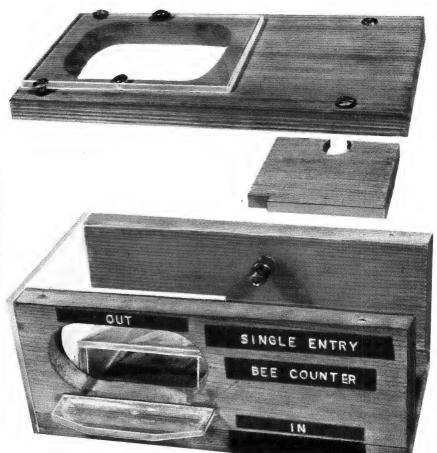
Connection to the battery is made via two wander plugs and a length of twin flex.

Fig. 3. Dimensions of the end made from Paxolin to accommodate the wander sockets for battery connection, and counterreadout.









A photograph of the prototype with top and tunnel lid (which holds PCC1 in position) removed. The photograph clearly shows the entrance and exit tunnels (labelled IN and OUT respectively). The take-off platform, made from Perspex, is located just beneath the exit cut-out, and is glued in position with Perspex adhesive.

EXIT AND ENTRANCE GEOMETRY

As said before, this device and its design utilises the bees' senses of smell and sight. From inside the hive, the exit from the hive appears as a bright opening to the outside world and so the exit path through the instrument must be a tunnel with transparent sides and top to allow this condition to be fulfilled.

In the instrument this tunnel slopes upwards so that when the bee emerges, it finds itself on a platform of Perspex, about ³4in. wide, situated above the hive base, and flies away.

When it returns, it will land on the hive base (landing/alighting board) and walk towards the hive.

The entrance to the hive is now through the Bee Counter which is a tunnel painted matt black; when the bee walks along the front of the instrument and reaches this tunnel it will enter.

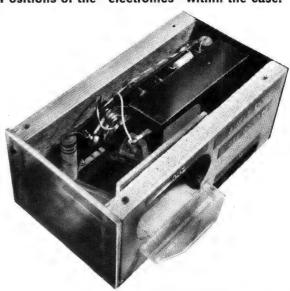
On entering, the tunnel becomes narrower and at the same time slopes upwards until it is just wide enough for a single bee to pass.

There is a lamp under the narrow part, with a hole in the floor of the tunnel, made up to the level of the floor with Perspex cement so that light can shine up through it.

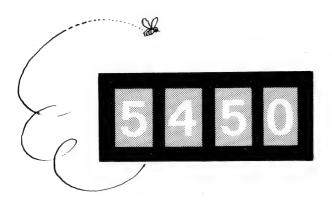
The light dependent resistor is situated in the

roof of the tunnel and as the bee walks between this and the lamp, the light beam is cut and the circuit activated.

Positions of the "electronics" within the case.



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CONSTRUCTION OF CASE

Cedar wood should be used to construct the case as this material will be readily acceptable to the bees.

Cedar wood will also withstand the weather without the need for painting but it is well to remember that if the counter is to be used in exposed outdoor conditions, weather protection becomes an important consideration, whereas in laboratory conditions it is not so.

The best compromise for an outdoor installation is a shelter which will keep off the rain.

First of all make all the wooden parts of the case as detailed in Fig. 4.

Now solder the two thin flexible covered wires to the bulb holder tags and screw in the bulb. These wires are led out through the top of the base and the bulb assembly is glued in position.

It is not likely that the bulb will need replacement because it is "under run" and there is a 20 ohm resistor (R1) in series with the bulb which reduces the light and heat dissipated in the bulb.

When the glue has set, fill up the light hole with Perspex cement so that it comes flush with the passage floor.

Glue down the two sides of the tunnel so that the width of the narrowest region is $^{1}_{4}$ in. Paint the tunnel top, bottom and sides a matt black.

The light dependent resistor should be a push fit into a hole in the tunnel roof.

Glue and screw the front and back to the base and glue the exit ramp in position. Drop the tunnel roof into position indicated. The other parts of the case are made from Perspex and their dimensions are given in Fig. 4.

With these made we can proceed with the assembly.

ASSEMBLY

Begin by screwing the Perspex side and top windows in position as indicated. Glue the Perspex platform to the front and place the Perspex exit guide in position.

Now solder the two wires from the bulb holder to the component board as detailed in Fig. 2, push the l.d.r. in position and then attach the board to the back of the case by means of a 4 B.A. nut and bolt. This bolt should be countersunk into the back so the back is flush with the front of the hive. If there is a gap here, the bees will try to go in or out through the smallest crevice.

Attach the wander sockets to the Paxolin side and solder to the appropriate flying leads from the component board. Next screw the Paxolin side to the case.

When the flying leads to the counter have been connected, fit the counter into its locating holes, (one end in the Paxolin and the other in the bracket on top of the Perspex exit guide) and secure with nuts. The counter digits should be visible through the slot in the Paxolin side.

Screw the top on and the unit is complete.

CAPACITY AND POSITION OF CASE

The single entry counter (as this is) is only suitable for a three or four frame hive, since with a full scale hive the returning bees would sometimes overload the tunnel capacity.

The maximum a single entry counter can handle is about 60 per minute.

For a full scale hive a three entry counter is necessary. This means the entry tunnel is divided into three passages, each with its own light beam arrangement, amplifier and counter.

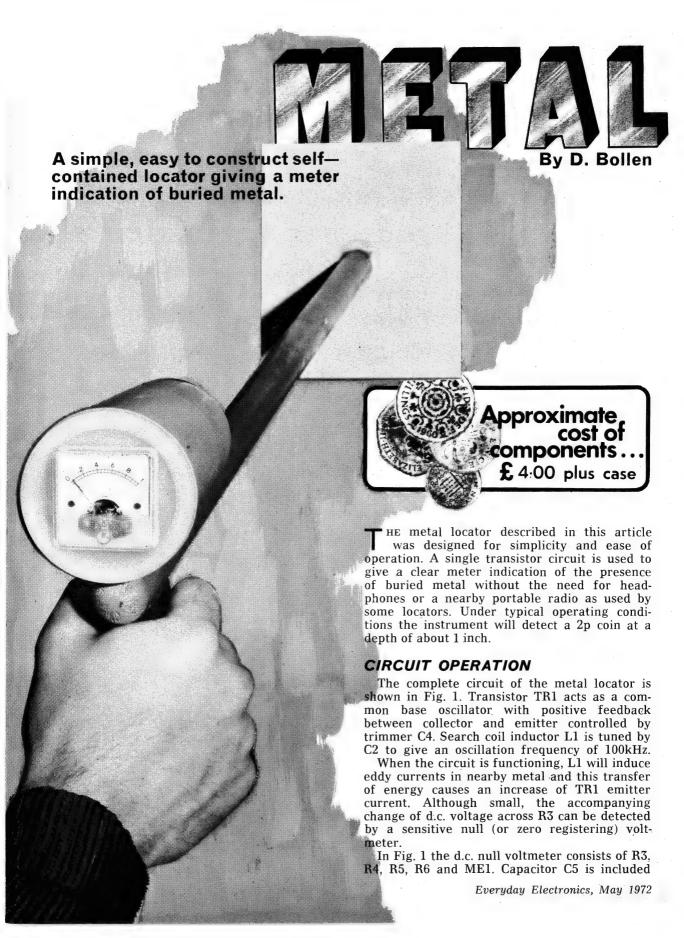
Whereas the single entry model is only 6¹4in. wide, which is about right for most observation hives, it is better to make the three entry model 16¹2in. wide so that it takes up the whole width of a Standard National hive.

When the counter is put in front of the hive the hive should be moved back by a distance equal to the depth of the Bee Counter, in this case 314 in. so that the point of entry is exactly as it was without the counter.

When this is done the bees will soon get used to the new conditions and will be using the exit and entry passages without any confusion.



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LOGATOB

to remove unwanted a.c. from the voltmeter input, and diodes D1 and D2 protect the meter movement against overload.

At a certain setting of C4, the d.c. voltage at TR1 emitter will equal the voltage at the junction of R5 and R6 so that no current flows through ME1; this can be taken as the normal operating point for the circuit. If metal is brought close to L1, the emitter voltage of TR1 will rise by several millivolts in relation to the voltage at the junction of R5 and R6, and the meter will read.

Full scale sensitivity of the null voltmeter is around 150 millivolts. Metal Locator response is shown in Fig. 2, where meter reading is plotted against depth for three weights of metal.

CONSTRUCTION

Commence construction by cutting a piece of $0\cdot 1$ inch matrix plain perforated circuit board to a size of $3\cdot 1$ by $1\cdot 4$ inches, and drill holes to take C4, VR1, and S1 (see Fig. 3).

Cut two brackets from a length of 12 inch aluminium angle and drill to accept the meter terminal screws and 6B.A. circuit board mounting screws.

Bolt the brackets to the circuit board, complete with solder tags, and insert all terminal pins in the positions shown in Fig. 3.

With C4, VR1, and S1 in place on the circuit board, proceed to mount and solder the remaining components in the following order; resistors, capacitors, wire links and leads, diodes and the transistor, using a heat shunt to protect the diodes and transistors while soldering them.

Obtain a plastic beaker with lid (of minimum dimensions 5 inches high by 2^{1}_{2} inches diameter) and cut away the centre of the lid to accept the meter MEI. Next, drill holes in the beaker for L1 leads, woodscrews, and to allow access to the circuit board controls, see Fig. 4.

When following the step-by-step instructions in Fig. 5, for making up the search coil L1, ensure that the pile windings can slide easily off the 5 inch diameter former. Short strips of insulating tape, placed sticky side out around the former, will hold the turns together and facilitate removal of the coil. Do not use Sellotape for this purpose as it is likely to damage the wire.

The metal locator frame (Fig. 4 and 6) consists of a chipboard or plywood handle, a $^{5}_{8}$ inch diameter dowel pole, and two s.r.b.p. or Perspex sheets for the search head. Screw and glue the handle to the pole and then glue the other end of the pole to the search head top board, this assembly can then be painted.

To complete the construction, screw the

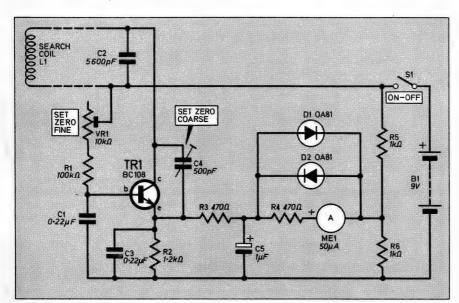


Fig. 1. Circuit diagram of the Metal Locator. The search coil L1 is mounted in the locator head and the dotted lines are the connecting wires to the circuitry.

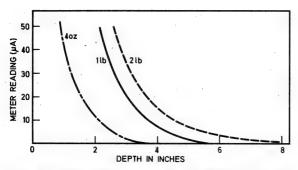


Fig. 2. Response curves of the Metal Locator.

plastic beaker to the pole opposite the handle, securely clamp the search coil between the boards, run twin leads from L1 to the beaker, and position the battery.

In the prototype, the battery was held in place behind the meter with a rubber band, as shown in the photograph, but it could equally well be fixed inside the beaker with a small clip or elastic band.

SETTING UP

Adjust VR1 to mid track, C4 to minimum capacitance (unscrewed), and switch on. The meter pointer should go beyond full scale. With the search coil well away from metal objects, screw in C4 until the meter reads somewhere between zero and full scale. Trim for a zero reading with VR1.

OPERATING LICENCE

The Metal Locator described in this article is designed to operate in the frequency band specified by the Ministry of Post and Telecommunications (16 to 150kHz). The circuit design of the locator should not be altered in any way that may affect the operating frequency.

A licence must be obtained before using the locator; this costs 75p for 5 years. An application form for a licence is obtainable from the Ministry of Post and Telecommunications, Waterloo Bridge House, Waterloo Road, London, S.E.1.

If the meter fails to read, or no response is obtained from adjustment of C4, check for wiring errors.

A certain amount of drift will be evident immediately after the locator has been switched on, therefore allow the circuit to settle down and then readjust C4 and VR1. Locator response can then be checked with metal weights and compared with Fig. 2.

Increased sensitivity can be achieved by reducing the value of C3 to 0.15μ F, but this will enhance circuit drift to the point where frequent adjustment of VR1 is necessary. Conversely, drift and sensitivity will be reduced if C3 is increased in value.

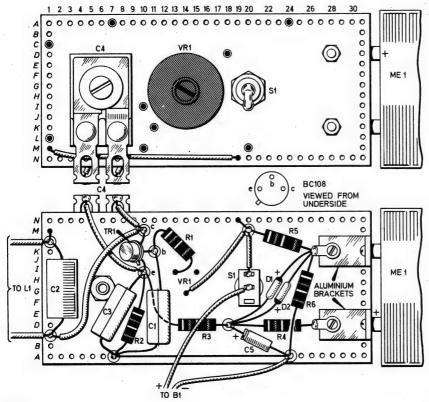
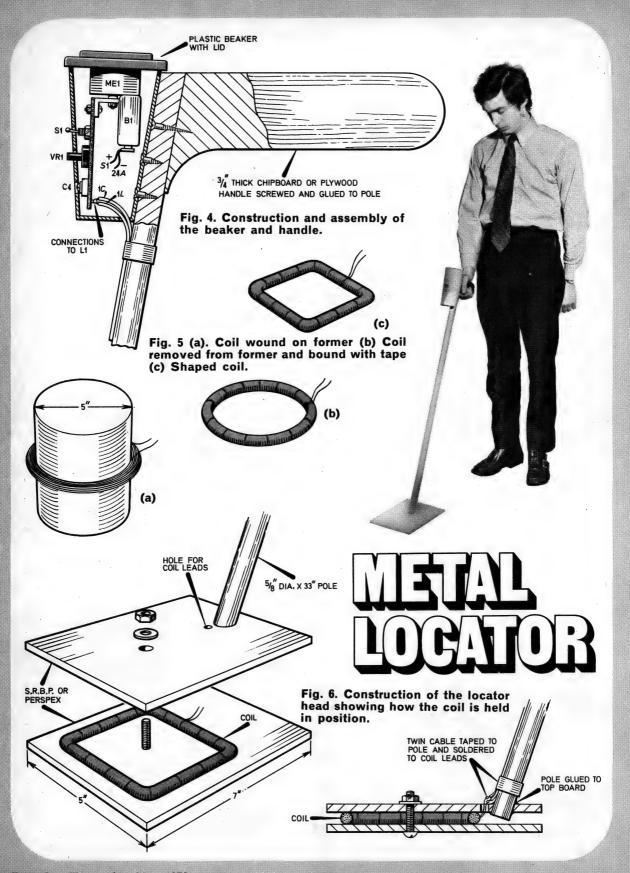


Fig. 3. Top and underside views of the circuit board and meter assembly. The circled connections represent the terminal pins used in the construction of this item. These pins are clearly indicated in the top diagram.



Components....

Resistors

R1 $100k\Omega$

R2 $1.2k\Omega$

R3 470Ω R4 470Ω

R5 $1k\Omega$

R6 $1k\Omega$

All \pm 10% $\frac{1}{2}$ watt carbon.

Capacitors

C1 0.22 µF polyester 250V

C2 5,600pF polystyrene

C3 0.22 µF polyester 250V

C4 500pF mica compression trimmer

C5 1μ F elect. 12V

Semiconductors

TR1 BC108 silicon npn

D1 OA81

D2 OA81

Meter

ME1 50μ A f.s.d. moving coil. SEW type

MR 38P

Switch

S1 S.P.S.T. sub-miniature toggle

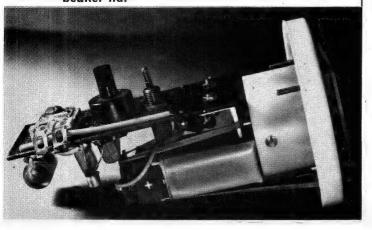
Miscellaneous

VR1 10k Ω miniature carbon T.V. type preset B1 PP3 battery. Circuit board 3·1 inch by 1·4 inch plain, perforated 0·1 inch matrix Veroboard and Veropins. 26 s.w.g. cotton covered or enamelled copper wire, plastic beaker (see text), connecting wire, wood and screws for assembly, $\frac{1}{2}$ in aluminium angle for brackets.

USE

The locator is now ready for use and can be used for beachcombing or searching the back garden or waste ground. The locator may be subjected to damp and the pole, in particular, should be painted for protection if nothing else.

Photograph showing the construction of the circuit board and meter mounted on the beaker lid.



Continued from page 361

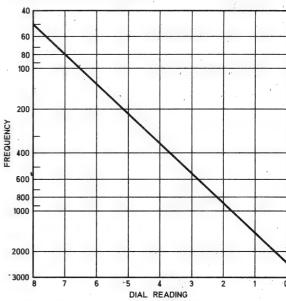


Fig. 8. Approximate output frequency for various control settings.

FINAL ASSEMBLY

Final assembly amounts to attaching the front panel to the box frame with self tapping screws, fitting the battery inside and fitting rear panel.

The generator can be connected to the input of any amplifier but the signal output level should be adjusted in accordance with that required by the amplifier input. To comply with the calibration chart given in Fig. 8 turn VR1 fully anti-clockwise and fix the frequency control knob to read zero. The output control knob is fixed in the same way i.e., to read zero with VR2 fully anti-clockwise.

The Audio Tone Generator is now ready for use and can be tried out in conjunction with a tape recorder.



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BRAND NEW GUARANTEED

LARGEST SELECTION OF SEMICONDUCTORS **COMPONENTS**

RETURN OF POST SERVICE

407	7117	71711										_
				TR/	NS	ISTO	RS				- 1	
10000	20p	2N3404	32½p	40310	45p	BC212L	18p	BSX28	32½p	NKT281	27½p	
2G301 2G302	20p 20p	2N3404 2N3405	45n	40311	35p	BCY30	27½p	BSX60	82 i p	NKT401	87 łp′ l	
2G303	20p	2N3414	901m	40312	47±D	BCY31	30p	BSX61	62 ł p	NKT402	90p	
2G306 2G308	42≟p 80p	2N3415 2N3416	221p 371p	40314 40320	371p	BCY32 BCY33	50p 25p	BSX76 BSX77	221p 271p	NKT403 NKT404	75p 62½p	
2G309	30p	2N3417	87 i p	40323	321 n	BCY34	30p	BSX78	27+P	NKT405	75p l	
2G371	15p	2N3570	£1-25	40324	471p	BCY38	40p	BSY10 BSY11	271p	NKT406 NKT451	62½p 62½p	
2G374 2G381	20p 22‡p	2N3572 2N3605	97≟p 27≟p	40326 40329	371p 30p	BCY39 BCY40	60p 50p	BSY24	271p 15p	NKT452	621p	
2N404	221p	2N3606	271p	40344	271p	BCY42	15p	BSY25	15p	NKT453	47+p	
2N696	90n	2N3607	221p 11p	40347	57±D	BCY43	15p	BSY26	171p	NKT6031	F821p	
2N697 2N698	17p 25p	2N3702 2N3703	11p	40348 40360	521p 421p	BCY54 BCY58	82½p 22½p	BSY27 BSY28	17±p	NKT6131 NKT6741	F 30n I	
2N706	124p	2N3704	11p	40361	47±p	BCY59	223p	BSY29	171p 171p	NKT6771	F 30n	
2N705A	TWEL	2N3705	10n	40362	571 p	BCY60	26.5 # D	BSY32	25p	NKT713	25p	
2N708 2N709	15p	2N3706 2N3707	09p 11p	40370 40406	821p	BCY70 BCY71	20p 25p	BSY36 BSY37	25p 25p	NKT781 NKT104	30p	
2N718	621p 25p	2N3708	07p	40407	571p	BCY72	17±p	B8Y38	221p	NKT104	39	
2N726	aup	2N3709	09p	40408	521p	BCZ10	271p	BSY39	221p	NKT105	87}p	
2N727 2N914	30p	2N3710 2N3711	09p 12p	40410 40467A	621p 571p	BCZ11 BD116	42}p £1·12}	BSY40 BSY51	32½p 32½p	MK1100.	32½p	
2N916	1710 1710	2N3715	£1.25	40468A	35p	BD121	65p	BSY52	821p 871p	NKT203	29	
2N918	800	2N3716	£1.30 £2.06	40600	571p 30p	BD123 BD124	82½p 60p	BSY53 BSY54	37}p 40p	NKT203	47½p	
2N929 2N930	221p 271p	2N3791 2N3819	35p	AC 107 AC126	20p	BD131	75p	BSY56	90p		371p	
2N1090	221p 221p	2N3823	97+p	AC127	25p	BD132	85p	BSY78	471p	NKT801	11	
2N1091	221p	2N3854	2/1P	AC128 AC154	20p	BDY10 BDY11	£1.87}	BSY79 BSY82	45p 52\p	NKT801	77½p	
2N1131 2N1132	25p 25p	2N3854A 2N3855		AC176	22 i p 25 p	BDY17	£1.50	BSY90	574P		97}p	
2N1302	174p	2N3855A	271p 30p	AC187	801n	BDY18	£1.75	BSY95A	124p	NKT801	13	
2N1303 2N1304		2N3856 2N3856A	80p 85p	AC188 ACY17	871p 271p	BDY19 BDY20	£1.971 £1.121	B8W41 B8W70	421p 271p	NKT802	£1.12	
2N1304	22 i p 22 i p	2N3858	25p	ACY18	25p	BDY38	97±p	Cili	75p		92½p	
2N1306	zop	2N3858A	a08	ACY19	25p	BDY60	£1.25	C424	27½p	NKT802	12	
2N1307 2N1308	25p 30p	2N3859 2N3859A	27 p 32 p	ACY20 ACY21	25p 25p	BDY61 BDY62	£1.25 £1.00	C425 C426	55p 40p	NKT802	92≟p 13	
2N1308	80p	2N3860	30p	ACY22	20p	BF115	25p	C428	371p 30p		921p	
2N1507	174p 25p	2N3866	£1.50	ACY28	20p	BF117	471p	C744		NKT802	14	
2N1613 2N1631	25p 35p	2N3877 2N3877A	40p	ACY40 ACY41	20p 25p	BF163 BF167	371p 18p	D16P1 D16P2	37½p 40p	NKT802	92½p	
2N1632	200	2N3900	371 p	ACY44	40p	BF173	19p	D16P3	371p		924p	
2N1638	274p	2N3900A	40p	AD140	521p	BF177	30p	D16P4 GET102	40p	NKT802		
2N1639 2N1671B	27 p	2N3901 2N3903	97≟p 35p	AD149 AD150	571p	BF178 BF179	30p 30p	GET113	30p 20p	OC20	92½p 75p	
2N1711	25p	2N3904	35p	AD161	621p 871p	BF180	35p	GET114	20p	OC22	50p	
2N1889	821p	2N3905	371p	AD162	871p	BF181	321p	GET118 GET119	20p 20p	OC23 OC24	60p 60p	ı
2N1893 2N2147	871p	2N3906 2N4058	87 ip 17 ip	AF106 AF114	421p 25p	BF184 BF185	25p 421p	GET120	521p	OC25	50n	ı
2N2148	57±0	2N4059	10n	AF115	25p	BF194	171p	GET873	12 i p 30 p	OC26	271p 621p	i
2N2160	571p 40p	2N4060	124p	AF116	25p	BF195	15p	GET880 GET887	30p 20p	OC28 OC29	621p 621p	i
2N2193 2N2193A	40p 421p	2N4061 2N4062	12ip 12ip	AF117 AF118	25p 621p	BF196 BF197	421p	GET889	221p	OC35	50p	ı
2N2194A	30p	2N4244	47 to	AF119	zup	BF198	423p	GET890	224p	OC36	621n	
2N2217	271p	2N4285	17 p	AF124 AF125	22½ p 20p	BF200 BF224	521p 14p	GET896 GET897	221p 221p	OC41 OC42	221p 25p	ı
2N2218 2N2219	28p 28p	2N4286 2N4287	17}p 17}p	AF125	20p	BF225	19p	GET898	22 p	OC44	20p	
2N2220	25p	2N4288	17∳p	AF127	174p	BF237	23p		E1-07	OC45	12}p	
2N2221 2N2222	25p 30p	2N4289 2N4290	17½p 17½p	AF139 AF178	87≟p 42≟p	BF238 BF244	23p 23p	MJ420 : MJ421 :	£1.12⅓ £1.12⅓	OC46 OC70	15p 15p	l
2N2270	47 1 p	2N4291	171p	AF179	721p	BFW61	471n	MJ430	1.02	OC71	12½p	l
2N2297	80p	2N4292	12}p	AF180	521 p	BFX12	221n	MJ440 MJ480	95p	OC72 OC74	124p	
2N2368 2N2369	171p 171p	2N4303 2N5027	471p	AF181 AF239	421p 421p	BFX13 BFX29	22 i p 30 p	MJ481	971p £1.25	OC75	32 ł p 22 ł p	l
2N2369A	174D	2N5028	571p	AF279	471n	BFX30	30p	MJ490	£1.00	OC76	22 tp	l
2N2410	421p	2N5029	471p	AF280	62 * p	BFX42	37 p 37 p	MJ491 MJ1800	21.37	OC77 OC81	30p 20p	ı
2N2483 2N2484	27½p 32½p	2N5030 2N5172	421p 121p	AF211 ASY26	32½p 25p	BFX44 BFX68	- 671p	MJE340	62 p	OC81D	22½n	l
2N2539	221 p	2N5174	92 ł D	ASY27	371p	BFX84	zop	MJE520	60p	OC83	25n	ı
2N2540	221p	2N5175	52½p	ASY28 ASY29	271p	BFX85 BFX86	32½p 25p	MJE521 MPF102	73p 421p	OC84 OC139	25p 321p	ı
2N2613 2N2614	85p 80p	2N5176 2N5232A	45p 30p	ASY36	271p 25p	BFX87	27±p	MPF103	37 p	OC140	321n	ı
2N2646	52 lp	2N5245	45p	ASY50	25p	BFX88	25p	MPF104	37}p	OC170	30p	ı
2N2696 2N2711	32½p 25p	2N5246 2N5249	42½p 67½p	ASY51 ASY54	32½p 25p	BFX89 BFX93	621p A 70p	MPF105 MPS3638	371p	OC171 OC200	30p 40p	ı
2N2711 2N2712	25p	2N5245	£3.25	ASY86	321p	BFY10	321n	NKT001	3 47 P	OC201	60p	ı
2N2713	27½p	2N5266	£2.75	AU103	£1.25	BFY11	42½p	NKT124	421p	OC202 OC203	75p	l
2N2714 2N2865	30p 62∤p	2N5267 2N5305	£2-621 371p	A8Z21 BC107	42 p 10 p	BFY17 BFY18	221p 321p	NKT125 NKT126	271p 271p	OC204	42½p 42½p	ı
2N2904	30p	2N5306	40p	BC108	10p	BFY19	82 p £1.60	NKT128	271p	OC205	BUD	ı
2N2904A	321 p	2N5307 2N5308	37±p 37±p	BC109 BC113	10p 15p	BFY20 BFY21	£1.60 42½p	NKT135 NKT137	27½p 32½p	OC207 OCP71	75p 421p	ı
2N2905\ 2N2905A	871p 40p	2N5308 2N5309	624p	BC115	15 p	BFY24	45p	NKT210	30p	ORP12	aup	l
2N2906	9.5 n	2N5310	4×+D	BC116A	15p	BFY25	25p	NKT211	30p	ORP61 P346A	50p	ı
2N2906A 2N2907	27 p 30p	2N5354 2N5355	27±p	BC118 BC121	10p 20p	BFY26 BFY29	20p 50p	NKT212 NKT213	30p	P346A TIS34	22½p 62½p	ı
2N2923	15p	2N5356	32 p	BC122	20p	BFY30	50p	NKT214	22½p	TIS43	27p	l
2N2924	15p	2N5365	471D	BC125 BC126	20p	BFY41	50p	NKT215	22 ł D	TIS44 TIS45	10p	ı
2N2925 2N2926	15p	2N5366 2N5367	32 i p 57 i p	BC120	20p 371p	BFY43 BFY50	62½p 23p	NKT216 NKT217	421p	TIS46	10p 11p	ı
Green	14p	2N5457	37±p	BC147	TOD	BFY51	ՁՈր	NKT219	30p	TIS47	11p	l
Yellow	12 p	28005	75p £2-00	BC148 BC149	10p 12p	BFY52 BFY53	23p	NKT223 NKT224	271p 25p	TIS48 TIS49	12½p 12½p	l
Orange 2N3011	12ip 80p	$28020 \\ 28102$	50p	BC152	171p	BFY56	171p A 571p	NKT225	22 i p	TIS50	171p	ł
2N3014	321p 18p	28103	25n	BC157	20n	BFY75		NKT229	30p	TIS51	17ip 12ip	ı
2N3053	18p	28104	25p	BC158	11p	BFY76 BFY77	42½p	NKT237	35p 25p	TIS52 TIS53	121p 221p	l
2N3054 2N3055	46p 62p	28501 28502	82⅓p 35p	BC159 BC160	12p 62}p 11p	BFY90	57½p 67½p	NKT238 NKT240	271p	TIS60	zz p	۱
2N3133	30p	28503	271p 40p	BC167	11p	BFW58		NKT241	271p	T1861 T1862	.25p	۱
2N3134 2N3135	30p 25p	3N83 3N128	40p 70p	BC168B BC168C	10p 11p	BFW59 BFW60	zop	NKT242 NKT243	20p 624p	T1862 T1P29A	271p 50p	۱
2N3135 2N3136	25p	3N140	771p	BC169B	11p	BPX25	£1.85	NKT244	174p	TIP30A	60n	۱
2N3390	25p	3N141	771p 721p	BC169C	12p	BPX29	£1.80	NKT245	20p	TIP31A TIP32A	621p	۱
2N3391 2N3391A	20p 80p	3N142 3N143	55p 67∔n	BC170 BC171	121p 15p	BPY10 BRY39	£1-45	NKT261 NKT262	20p 30p	TIP33A	75p	۱
2N 3391A	17}p	3N152	67±p 87±p	BC172	15p	B8X19	37½p 17½p	NKT264	20p	£	1.021p	۱
2N3393	15p	R.C.A.	52½p	BC175	221p 10p	B8X20 B8X21	17+p	NKT271 NKT272	20p 20p	TIP34A TIP35A	£2·05	۱
2N3394 2N3402	15p 22‡p	40050 40251	55p 32⅓p	BC182 BC183	09ъ	BSX26	87±p 45p	NKT274	20p	TIP36A		۱
2N3403	zz i b	40309	32}p	BC184	11p	B8X27	471p	NKT275	20p			۱
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SN7407	0.80	0.75	SN7443	2.86	2.70	SN7481	1.40	
SN7408	0.20	0.18	SN7444	2.86	2.70	SN7482	0.87	
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18113	15p	BA100	15p	BY103	22p	OA70	7p
18120	12p	BA102	25p	BY122	471p	OA73	10p
18121	14p	BA110	25p	BY124	15p	OA79	7p
IS130	8p	BA114	15p	BY126	15p	OA81	8r
18131	10p	BA115	7p	BY127	17p	OA85	10p
18132	12p	BA141	17p	BY164	57p	OA90	7p"
18920	7p	BA142	17p	BYX10	22 p	OA91	7p
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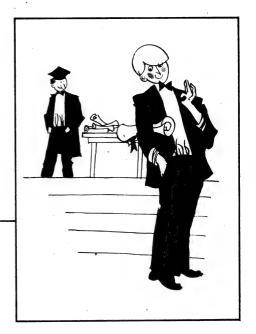
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Last month we posed some problems under the heading Teach in Half-Term Test. We will now answer those problems and try to show how we arrived at the answers. If you have got some of them wrong do not worry, just try and follow our explanation and see where you went wrong.

(1) They flow from negative to positive in reality. Although we assume that conventional current flows from positive to negative the actual electrons flow from negative to positive.

(2) (b) μ A (microamps), (e) A (amps) (3) 22 volts. V = IR hence V = 0.01 \times 2.2 \times 1,000 = 22V (4) It does not matter. All the resistor does is to limit the current; this can be done at any point around

the circuit.

Total current
$$I \Rightarrow \frac{V}{D} = \frac{9}{42.3} = 0.21 A$$

the circuit. (5) 2-8mA. Total resistance is $2\cdot 2k\Omega+1k\Omega=3\cdot 2k\Omega$. Current flow I = $\frac{V}{R}=\frac{9}{3\cdot 2\times 1,000}=2\cdot 8-1,000$ A= $2\cdot 8m$ A (6) R1 and R3 $\frac{1}{2}$ W, R2 1W. Total circuit resistance $R_T=R1+\frac{R2\times R3}{R2+R3}=10+33\cdot 3=43\cdot 3\Omega.$ Total current I = $\frac{V}{R}=\frac{9}{43\cdot 3}=0\cdot 21$ A Dissipation of R1 = $1^2R=0\cdot 21\times 0\cdot 21\times 10=0\cdot 44$ W. The nearest commercial rating is $\frac{1}{2}$ W. Next calculate the voltage drop across R2 and R3 together V = IR = $0\cdot 21\times 3\cdot 3\cdot 3=7$ V. 33.3 = 7V.

We know that $W = I^2R$, but $I = \frac{V}{R}$ therefore

$$W = \frac{V}{R} \times \frac{V}{R} \times R$$
 and, cancelling $W = \frac{V^2}{R}$

Dissipation in R2 =
$$\frac{V^2}{R} = \frac{7 \times 7}{50} = \frac{49}{50} = 0.98W$$

Dissipation in R3 = $\frac{V^2}{R} = \frac{49}{100} = 0.49W$

Dissipation in R3 =
$$\frac{V^2}{R} = \frac{49}{100} = 0.49W$$

(7) 0.4W or 400mW. Maximum dissipation occurs when the value of VR1 equals that of R1 i.e. 50Ω . When both resistors are of equal value the voltage drop across each is half the voltage drop across both,

therefore, maximum dissipation in VR1
$$= \frac{V^2}{R} = \frac{4.5 \times 4.5}{50} = \frac{20.25}{50} = 0.405W$$

(8) (a) $4.7k\Omega \pm 10\%$

(b) $22k\Omega \pm 5\%$

(c) $100k\Omega \pm 10\%$

(9) (b) $20\mu F$ 40V. In most applications using electrolytic capacitors the capacitance must be greater than a certain value; the tolerance of a normal 16µF would encompass 20µF. The important thing is that the working voltage is the same or greater.

(10) Reject it politely. He has given you a 120,000pF or 0·12μF capacitor. Check to see if he has the precise value and, if he does not, you may as well take this one, since it should be near enough to use as a substitute.

(11) C1 will charge up the fastest as it has the lowest value and is being charged through the lowest value resistor.

(12) C2 will take the longest time to charge, as it has the highest value and is being charged through the highest value resistor.

(13) Forward biassed. The conventional current flows from positive to negative and can thus flow through the diode in the direction of the arrow.

(14) 100V and 100mA. Peak reverse breakdown voltage will be the battery voltage. Since in the reversed biassed condition there is negligible current flowing R1 will not drop any voltage and the full supply voltage will appear across D1. In the forward biassed condition the diode can be assumed to be a short circuit thus only R1 can limit the current flowing hence

$$I = \frac{V}{R} = \frac{100}{1 \times 1,000} = 0.1 A \text{ or } 100 \text{mA}$$

 $I = \frac{V}{R} = \frac{100}{1 \times 1,000} = 0.1 \text{A or } 100 \text{mA}$ (15) (d) 100V, 150mA. Both ratings given are minimum ratings, 0.1A = 100mA.

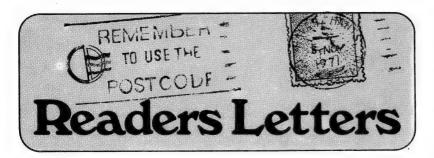
(16) (b) 0.6V. As the diode is forward biassed the voltage would be 0.6V. There is always a voltage drop of approximately 600mV across silicon diodes due to the "knee" in the characteristic.

Well, how did you fare? If you got them all right that is excellent, if you did not the important thing is that you understand where you had difficulties. We suggest that you re-read the relevant sections of the Teach-In series.

We hope that you found the questions a challenge and at the same time they have opened your eyes to some calculation methods-particularly the calculation of dis-

sipation. If you used $W = I^2R$ instead of deriving $W = \frac{V}{R}$

this does not matter but it may pay to look for an easier way next time.



Bias Value

Having been a subscriber to *P.E.* and *P.W.* "off and on" for about 10 years I came across the January issue of EVERYDAY ELECTRONICS, which had my instant approval and now joins the rank of my other magazine's culminating in an endless and very informative pile on top the piano.

I find it is a magazine not only of theoretical enthusiasm but of great practical interest to the "everyday handyman" and certain to be a book for beginners, especially the very helpful facts "projected" by Mike Hughes, M.A.

I would hope in the future that perhaps Mr. Hughes could give reference to finding values of bias resistors, etc., needed for the satisfactory operation of different transistor parameters, and also relevant circuit operation of thyristors, unijunction and field effect transistors and other very useful flexible types of semiconductors.

Noticing other readers' troubles referring to the *Electro Laugh*, I also constructed this article and it worked first time owing to the way I adopt when working on, or constructing any project, I always check the finished article with the actual circuit diagram thus finding our little friend Q7 and P7.

Unfortunately the only earphone I had was a high impedance crystal type, but by connecting a resistor in the region of 250 ohms in parallel with it, it brought the overall impedance down to a satisfactory level with a slight reduction in volume.

J. Mason S. Wales

We doubt if Teach-In will be able to meet all your needs as it will finish after 12 months. However we will be publishing further series that should help.

Another Bug

Naturally, I was quite flattered to discover that you had found my letter sufficiently interesting for inclusion in *Readers Letters* (March issue), however, I must admit that my pleasure was mixed with large helpings of disappointment and frustration due to your editing of the letter.

I am not complaining at all about the amount of space allocated to my comments—I realise you have the right to include only that which in your wisdom you decide is worthy of publication.

My complaint is that you have entirely neglected to make even a brief reference to what was after all the main point of my letter-the difficulty of obtaining items advertised in your magazine. By omitting any reference to this frustrating situation, my letter as printed is sailing under false colours-the few minor constructional queries were in fact, sorted out by trial and error once I got going. The real reason for being unable to get cracking was not so much mounting components, as actually getting hold of them!

The fact that you completely ignored my comments regarding suppliers leads me to two conclusions:

(One) That you accepted my comments to be an exaggeration of a somewhat hysterical nature, and were not a true picture of the real situation, or

(Two) That you accepted my statements as correct, but did not wish to offend your advertisers whose business you must obviously wish to retain.

With regard to the former, I feel I, must now justify my remarks by quoting a few of the more deplorable examples of SERVICE, and leave you to form your own conclusions. These examples are on a separate sheet herewith enclosed.

Regarding (Two), whilst I

realise that you are not to be held responsible for goods or services advertised in your columns, you do, however, have a moral responsibility to your readers. After all, it is you that place these offers before us, the readers, and if for example, I had not seen a certain item offered in your magazine, then I would have been saved the trouble and frustration that followed when the item failed to arrive, and all attempts to obtain satisfaction are largely ignored.

However, I have now found a couple of very good suppliers whose friendly, courteous, and extremely efficient service have allowed me to obtain some of the pleasure that I had hoped would be derived from my new hobby (Galleon Trading Co. and Radio Exchange Co.).

To date I have completed several very efficient radios, some from kits; also the *Astron*, a general purpose amplifier, and one or two other gadgets, and success rate so far is quite satisfactory, so the situation is not too

black after all.

J. G. Richards Sale, Cheshire

The above correspondent supplied us with details of orders placed with four different advertisers, none of which had been expediently dealt with, at the time of writing.

We have investigated all of these cases on behalf of our reader. The delays, regrettable as they are, seem to be unavoidable and can be largely attributed to the phenomenal success of this magazine's declared intention to popularise the hobby of electronics!

As a consequence, our advertisers are sometimes overwhelmed by a flood of orders, and delays do therefore sometimes arise. But we know all our advertisers make determined efforts to clear their back-log of orders as quickly as possible.

We, on our part, will always investigate any serious and reasonable complaints, on behalf of our readers.

Cell Life

I have just read the March issue of EVERYDAY ELECTRONICS and thoroughly appreciated the Ruminations by Sensor where he mentioned the tin saw and how much damage could result to a

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Designed to operate transistor sets and amplifiers. Adjustable output 6v., 9v., 12 volts for up to 500mA (class B working). Takes the place of any of the following batteries: PP1, PP3, PP4, PP6, PP7, PP9 and others. Kit comprises: mains transformer rectifier, smoothing and load resistor, condensers and instructions. Real snip at only 35p, plus 20p postage.

MICRO SWITCH 5 amp. changeover contacts, 9p each, \$1 doz. 15 amp. Model 10p each or \$1.05 doz.





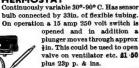
EXTRACTOR FAN EXTRACTOR FAN
Cleans the air at the rate of
10,000 cubic ft. per hour.
Suitable for kitchens, bathrooms, factories, changing
rooms, etc., it's so quiet it can
hardly be heard. Compact, 5½
casing with 5½ fan blades.
Kit comprises motor, fan
blades, abeet steel casing, pull
switch, mains connector, and
fixing brackets, £2 plus 36p
post and ins.

MAINS MOTOR



Precision made—as used in record decks and tape recorders-ideal also for extractor fan, blower, heaters, etc. New and perfect. Snip at 50p. Postage 15p for first one then 5p for each one ordered.

THERMOSTAT



5A 3-PIN SWITCHED SOCKETS

An excellent opportunity to make that bench dis board you have needed or to stock up for future jobs. This month we offer 6 British made (Hieratt) bakelite flush mounting shuttered switch sockets for only 80p plus 18 poet and insurance. (20 boxes post free).



MAINS OPERATED

MAINS OPERATED SOLENOIDS
Model 772—small but powerful 1' pull—approx. size
1 12' 809.
24' x 2' 11' 809.
Model TT10 11' pull. Size
24' x 2' 11' 81' 81' 81' 80
plus 20p post and insurance.

4 TELESCOPIC AERIAL

E

for portable, car radio or transmitter. Chrome plated—
tr

3 STAGE PERMEABILITY TUNER



This Tuner is a precision instrument made by the famous ment made by the famous famous Radiomobile Car Radio. It is a medium wave tuner (but set of longwave coils available 25p) with a frequency coverage 1620 Ke/s-525 Ke/s and intended to operate with an I.F. value of 470 Ke/s. Extremely compact (size only 2½ × 2 × jns. thick) with reduction gear for fine tuning. Snip price this month 50p, with circuit of front end suitable for car radio or as a general purpose tuner for use with Amplifier. Post free.

CAPACITOR DISCHARGE CAR IGNITION

This system which has proved to be amazingly efficient. We offer kit of parts as FW circuit £5.95 plus 20p p. é. p. De-luxe model with prepared circuit board £6.95. When ordering please state whether for positive or negative systems.

RADIO STETHOSCOPE

ELECTRONIC IGNITION

Easiest way to fault find—traces signal from aerial to speaker—when signal stops you've found the fault. Use it on Radio, TV, amplifier, anything—complete kit comprises two special transistors and all parts including probe tube and crystal earpiece. 28—twin stethoset instead of earpiece 75p extra post and in a 20n. set instead of earpiece extra post and ins. 20p.



STANDARD WAFER SWITCHES

Standard size 11" wafer-silver-plated 5-amp contact, standard ?" spindle 2" long-with locking washer and nut. No. of Poles 2 way 3 way 4 way 5 way 6 way 8 way 9 way 10way 12way
 7 Sway 6 way 8 way 9 way 10way 12way

 40p
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9 poles 10 poles 11 poles 12 poles

THYRISTOR LIGHT DIMMER

For any lamp up to 200 watt. Mounted on switch plate to fit in place of standard switch. Virtually no radio interferences. Price \$1.99 plus 20p post and insurance.



- THIS MONTHS SNIP



1 HOUR MINUTE TIMER, Made by Smiths complete with control knob and calibrated dial. This month's special bargain at 50p. Useful in the Kitchen, Office and Dark-room etc.

MULLARD AUDIO AMP LIFIER MODULE



Uses 4 transistors, and has an output of 750m W into 8 ohms speakers. Input suitable for crystal mic. or pick-up. 9 voit battery operated. Size 2" long × 1½" wide × 1" high. SPECIAL SNIP PRICE 60p each. 10 for \$5.

POCKET CIRCUIT TESTER

Test continuity for any low resistance circuit, house wiring, car electrics. Test polarity of diodes and rectifiers. Also ideal size for conversion to signal injector (circuit supplied), 30p or 2 for 50p. Post paid.



METAL LOCATOR AUDIO TONE GENERATOR **BEE COUNTER**

To receive details on these kits send s.a.e. for parts list.

MULLARD I.F. MODULE

This is a fully screened intermediate frequency module for amplification and detection of f.m. signals at 10-7MHz and a.m. signals at 470-MHz. The first stage is used as an i.f. amplifier for f.m. and a self oscillating mixer for a.m. operation, in conjunction with an external oscillator coil. 75p each. 10 for 28-75.

DISTRIBUTION PANELS

4 4 4 4 Just what you need for work bench or lab.

4 × 13 amp sockets in metal box to take
standard 13 amp fused plugs and on/off switch with neon warning light. Supplied
complete with 6 feet of flex cable. Wired up ready to work.

BATTERY CONDITION TESTER

Made by Mallory but suitable for all batteries made by Ever Ready and others, most of which are zinc carbon types but also farcury manganese—nicad—silver oxide and alkaline batteries may be tested. The tester puts a dummy load on the battery and the meter scale indicates the condition depending upon which section the pointer rests. The section reads "replace" "weak" or "good". The tester is complete in its case, size 3½ x 6½ x 2° with leads and prods. Price 41-75 plus 20p postage.

Thermostat with Probe. Made by the famous Ranco Thermostat Co. Covers the range from approx. 0°-20°CC. variable by a control spindle, handles currents up to 16 amps. Length of capillary and sensor tube approx. 3° 6°. These are ideal for ovens and as a general purpose thermostat. Price 50p each or 10 for \$4.50. Small Tuning Condenser as fitted to many imported Japanese and Hong Kong radios. 2 gang about 2007F per gang. Size approx. 1° × 1° × 1° × 50p each or 10 for 45p. Speciate of 10 for 45p. Speciate for 10 for 45p. Speciate Frames. (No lenses) with dust cover. 25p each or 10 for 45p. Speciate Frames. (No lenses) with built-in hearing aids. The amplifier and battery being housed in the arms. Although these are complete hearing sids we are selling them purely for the sub miniature components they contain. We give no guarantee that they are in working order also these may be secondhand. Price 20.50 each. Foot Switch, Twin levers each of which operates a 10 amp QMB changeover switch. Price 90p each. Programmers. 5 Revs per minute. Made by Magnetic Devices Ltd. The contacts may be set to trigger anywhere around the shatt, ideal for motivated lighting displays, sequential switching etc. Drive motors are 200–240° 50Hz. Model A has 5 change over contacts. Price \$3.50. Model B has 11 change over contacts. Price \$3.50. Model B has 11 change over contacts. Price \$3.8 Radiant Cooker Rings. As fitted to Tricity and many other popular cookers. We have two types. These are copper clad \$' tubular construction. Both models having an external diameter of 6\$' and the elements have been slightly flattened to increase radiation.

to increase radiation.

Backer Model 7DI MkII again 2000 watts rated but 230-240v, has no cover over element ends. Price 65p each or 10 for 25-85.

Slide Switch. 2 pole change over panel mounting by two 6, BA screws. Size approx. 1° × §* rated 250v lamp, 6p each. 10 for 54p. 100 for 25-10. 500

each. 10 for \$4p. 100 for \$2.10. 500 . Left for \$24. As above but for printed circuit 5p each, 10 for 45p. 100 for \$4.25. Sub Miniature Slide Switch. DPDT 19mm (\$*approx.) between fixing centres. 12p each or 10 for \$1.08.

KITS FOR PREVIOUS PROJECTS

Unless otherwise stated, kits contain electronic parts only. The case and special items can be obtained locally. Also batteries are not included. Kits may be returned for refund if construction has not been started. We reserve the right to substitute components should deliveries be protracted so as to avoid undue delay.

HOME SENTINEL INTRUDER ALARM
Electronic Components with suitable case
23.75
SNAP INDICATOR
WINDSCREEN WIPER CONTROL
Components including metal for chassis #2
RECORD PLAYER, All components, but not
case, loudspeaker, record deck or pick-up
\$5.50
DEMO DECK 26.75 POST PAID
FUZZ BOX 21-85
PHOTOGRAPHIC COLOUR
TEMPERATURE METER
ASTRON RADIO
REMOTE TEMPERATURE
COMPARATOR£4.25
ELECTRO LAUGH
TRANSISTOR MICROPHONE \$1.70
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All electronic parts and metal bracket £2.50
RAIN WARNING ALARM
All electronic parts and chassis \$1.80
WA-WA PEDAL
DARKROOM TIMER
SIGNAL INJECTOR80p
SOIL MOISTURE METER
SIMPLE CALCULATOR£2-20
D.C. POWER SUPPLY 25-00
BABY ALARM£4.00

Mains Transformer, Primary 240v. tapped 220v. Secondary 20v. ; amp. Price 60p each or 10 for 25-40.

for \$5-40. Dial Thermometer—reading from 200-525F used on Tricity and other cookers. This has a flange and can be mounted through a 1½ hole or alternatively it can just be rested on the object whose temperature it is required to measure. Size 2 × ½ overall diameter. Depth ½ below and ½ above mounting panel. Price 80p each or 10 for \$7.20p.

24-HOUR TIME

24-HOUR TIME
WITCH
Made by Smiths, these are
AC mains operated, NOT
CLOCKWORK. Ideal for
mounting on rack or shelf or
can be built into box
with 13A socket. 2 comperiods per 24 hours, 5 amp changeover contacts
will switch circuit on or off during these periods.
29:40 post and ins. 23p. Additional time contacts
50p pair.

Where postage is not stated then orders over 25 are post free. Below 25 add 20p. Semiconductors add 5p post. Over £1 post free. S.A.E. with enquiries please.

J. BULL (ELECTRICAL) LTD.

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HENRY'S LOW COST FIRST GRADE BRANDED GERMANUM and SILICON TRANSISTORS, DIODES, RECTIFIERS, DIODES, DIO BY ATES EMIHUS FAIRCHILD FERRANTI I.T.T. MULLARD NEWMARKET PHILIPS R.C.A. TEXAS

TRANSISTORS

A SELECTION FROM OUR LIST					
AAY30 AAY42	10p 15p	BD115 75p BD123 85p	OC16 50p OC20 85p	2N1303 18p 2N1304 22p	
AAZ13	10p	BD124 80p		2N1304 22p 2N1305 22p	
AAZ15 AAZ17	10p 10p	BD131 75p BD132 80p	OC23 60p OC24 60p	2N1306 25p	
AC107 AC126	35p 30p	BDY 11 £1.50	OC25 40p OC26 25p	2N1307 25p 2N1308 25p	
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AC127Z AC128	50p 20p	£1.50 BDY38 65p	OC29 60p OC35 50p	2N1613 20p 2N2147 75p	
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AC188	25p		OC43 50p	2N2218 20 p	
ACY17 ACY18	30p 25p	BF154 20p	OC44 15p OC45 12p OC70 12p	2N2218A 25p	
ACY19 ACY20	25p 20p	BF158 15p BF159 35p	OC71 12p	2N2219 20p 2N2219A	
ACY21	ZUD	BF180 35p	IOC/2 2001	95n	
ACY22 ACY39	10p 55p	BF195 15p	OC76 25p	2N2221 200	
ACY40 ACY41	20p	BF196 15p BF197 15p	OC81 20p	2N2221A 25p	
ACY44 AD140	25p 50p	REV12 05m	OC81D 20p OC81Z 40p	2N2222 20p 2N2222A	
AD149	500 l	BFX 30 25n	OC83 95n	25p	
AD161 AD162	35p 35p	BFX84 25p BFX85 30p	OC84 25p OC139 25p	2N2369 15p 2N2369A	
AF114 AF115	25p 25p	BFX86 25p	OC140 40p OC141 60p	2N2646 40p	
AF116	96-1	BFX88 20p	OC170 25p	2N2904 20p	
AF117 AF118	20p 60p	BFY50 20p	OC171 80p OC200 40p	2N2904A 25p	
AF124 AF125	25p 20p	BFY51 20p BFY52 20p	OC201 75p	2N2905 25p 2N2905A	
AF126	zun i	BFY53 15p	OC206 95p	25p	
AF127 AF139	20p 30p	BSX20 15p	OCP71 97p ORP12 50p	2N2906 20p 2N2906A	
AF180 AF181	50p	BSV97 15n	ORP60 40p ST140 15p	25p 2N2907 23p	
AF185 AF186	50p 40p	BST95A 12p BSY95 12p	ST141 20p TIP29A 50p	2N2907A	
AF239	40p i	BU105 £2.25	TTP30A 60 n	25p 2N2925 15p	
ASY26 ASY27	25p 30p	BY100 15p BY126 12p	TIP31A 60p TIP32A 70p	2N2926 10p 2N3011 20p	
ASY28 ASY29	25p 80p	BV194 15n	TIP33A £1.00	2N3053 20n	
ASZ21	55p	BYZIU BADI	TIP34A	2N3055 75p	
BA100 BA102	15p 30p	BYZ11 30p BYZ12 30p	£1.50 TIS43 40p	2N3525 £1.00	
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BAX13	5p	GET102 35p	TIS52 10p	2N3703 10p	
BAX16 BAY31	7p 7p	GET111 45p	TIS60 18p TIS61 20p	2N3704 12p 2N3705 10p	
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BC119 BC134	20p	MJ420 80p MJ421 80p	ZTX 504 25p	2N3773 £2·35	
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BC127	20p 20p	#1·10 MJ2901	ZTX 503 17p	£2·75 2N3819 35p	
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BC148 BC149	10p 12p	MJE370 20p MJE371 80p	IN914 7p IN916 10p	2N3824 95p 2N3903 20p	
BC153	20p	MJE520 75n	IN4148 7p	2N3906 25n	
BC154 BC167	20p 15p	MJE521 70p MJE2955	I844 7p I8921 7p	2N4058 12p 2N4061 12p	
BC169 BC169C	12p 15p	£1·10 MJE3055	18922 8p 2G301 25p	2N4062 12p 2N4289 12p	
BC177	200	75p MPSA06 25p	1 2G302 80p	2N4290 12m	
BC178 BC179	20p 20p	MPSA56 20n	2N404 20p	2N4871 35p 2N5457 30p	
BC182L BC183L	10p	MPSA70 15p MPSU06 50p	2N696 15p 2N697 15p	2N5458 35p 2N5459 40p	
BC184L BC212L	12p	MPSU56 55p	2N698 30p	40250 50p	
BC212L	12p	NKT216 35p	2N706A 12p	40361 40p	
BC214L BCY30	15p 35p	NKT403 75p NKT404 55p	2N708 15p 2N930 20p	40302 300	
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BCY34 BCY38	35p 45p	OA70 10p OA79 10p	2N1302 18p	40636 \$1.10	
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HENRY'S LOW INTEGRATED CIRCUITS

BRAND NEW FULL SPECIFICATION TTL74 SERIES BRANDED FAIRCHILD, I.T.T. AND TEXAS

	DESCRIPTION OF STREET OF STATE		OTTAN	MINTEN	TOWN	
	DEVICES MAY BE MIXED TO QUAL	TEX LOS	LAUAN	TITE PR	LIUES	250 +
No.	Description	1-11	12-24	25-99	100+	
7400	Quadruple 2-input NAND gates	20p	18p	14p	14p	12p
7401	Quad 2-input open collector NAND gates	20p	18p	16p	14p	12p
7402	Quad 2-input NOR gates	20p	18p	16p	14p	12p
7403	Quad 2-input open collector NAND gates	20p	18p	16p	14p	12p
7404	Hextuple inverters	20p	18p	16p	14p	12p
7405	Hex inverters with open collector outputs	20p	18p	16p	140	12p
7410	Triple 3-input NAND gates	20p	18p	16p	14p	12p
7413		30p	27p	25p	22p	20p
	Dual 4-input Schmitt triggers	20p	18p	16p	14p	12p
7420	Dual 4-input NAND gates				14p	12p
7430	Single 8-input NAND gates	20p	18p	16p		
7440	Dual 4-input NAND buffer gates	20p	18p	16p	14p	12p
7441	BCD-Decimal decoder/Nixie driver	75p	72 p	70p	60p	55p
7442	BCD-Decimal decoder (4-10-line) TTL O/P	75p	72p	70p	60p	55p
7443	Excess 3-Decimal decoder TTL outputs	£1 · 00	95p	90p	80p	70p
7447	BCD-Decimal 7 seg. decoder/indicator driver	£1 · 75	£1 · 60	£1 · 45	£1·30	£1 · 18
7448	BCD-Decimal 7 seg, decoder/driver TTL O/P	£1.75	£1.60	£1 · 45	£1·30	£1 · 1
7450	Expand dual 2-input AND-OR-INVERT gates	20p	18p	16p	14p	12p
7451	Dual 2-wide 2-input AND-OR-INVERT gates	20p	18p	16p	14p	12p
		20p	18p	16p	14p	12p
7453	Quad 2-input expand AND-OR-INVERT gates		18p	16p	14p	12p
7454	4-wide 2-input AND-OR-INVERT gates	20p				12p
7460	Dual 4-input expanders	20p	18p	16p	14p	
7470	Single J-K flip-flop (gated inputs)	30p	27p	25p	22p	20p
7472	Single J-K flip flop (gated inputs)	30p	27p	25p	22p	20p
7473	Dual J-K flip flop	40p	87p	35p	33p	30p
7474	Dual J-K flip flop	40p	37p	35p	38p	30p
7475	Quadruple bistable latch	45p	42p	40p	38p	35p
7476	Dual J-K flip-flops with Preset and Clear	40p	37p	34p	31p	28p
	Coted Pull Adder	80p	72p	67p	59p	55p
7480	Gated Full Adder 16-bit read/write memory 2-bit binary Full Adder 4-bit binary Full Adder	£1.25	£1.15	£1.10	£1.00	90p
7481	16-bit read/write memory		80p	70p	65p	60p
7482	2-bit binary Full Adder	87p				
7483	4-bit binary Full Adder	£1 · 00	90p	85p	80p	73p
7484	4-bit binary Full Adder 16-bit RAN with gated write inputs	90p	85p	80p	75p	71p
7486	Quadruple 2-input Exclusive OR gates	45p	41p	38p	35p	33p
7490	BCD decade counter	75p	70p	65p	60p	55p
7491	8-bit shift register	£1.00	95p	90p	80p	70p
7492	Divide twelve counter	75p	70p	65p	60p	55p
7493	4-bit bingry counter	75p	70p	65p	60p	55p
7494	Deel ontwo 4 bit shift register	80p	75p	70p	65p	60p
	A hit are down white register	80p	75p	70p	65p	60p
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7496	A-bit binary counter Dual entry 4-bit shift register 4-bit up-down shift register 5-bit parallel/serial in/out shift register 8-bit bistable latch Hextuple Set-Reset latches			£2 00	£1 · 75	£1 · 50
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74197	Asynchronous presettable 4-bit binary counter	£1.50	£1 · 40	£1.30	£1·10	£1 · 00
Compl	ete data on the above in booklet 20 pages, Ref. 29,	issue 2 at				
Town	I.C. Handbook. Complete information on 10	0 types	ton. Pos	t 10p.		
TCYSTS	1.C. Handbook, Complete mormation on 10	T 90n		P*		

Integrated circuit sockets 14 pin D.I.L. 25p; 16 pin D.I.L. 30p

INTEGRATED CIRCUITS MFC4000P 55p 60p £2:50 £1:50 £1:50 £1:50 MFC4000P MFC4010P 1C12 PA246 TAD100 TAD110 TAD110 MC724P 702C (TO5) 709C (TO5) 709C (D.I.L.) 723C(TO5) 741C(TO5) MC1303P MC1304P 50p 75p 45p 45p 45p £1.00 80p £2.00 £2.25 £1.50 SL403D 741C(DIL)

914(TO5) 923(TO5) TOSHIBA

20 watt amp. TOSHIBA

PLESSEY INTEGRATED CIRCUIT 3 Watt Amplifier 8L403D Complete with 8-page booklet, circuits S. C. V £1.50

ZENER DIODES 12p 10p 8p 7p 6p series*
1½ watt
ZL series 25p 23p 20p 17p 15p 3 watt 3TZ series 30p 27p 25p 22p 20p 10 watt ZS series +40p 37p 35p 30p 25p 10 watt
ZS series +40p 37p 35p 30p 25p
All types are 5%. Wire Ended + these
Stud. in voltages 3·3-100 volt in all
standard values. *3·3-33 volt.

SILICON RECTIFIERS

1 AMP MINIATURE PLASTIC WIRE ENDED

Type P.I.V. 1-49 50+ 100+ 500+1000+
1N4001 50 69 59 419 45 31n 0+ 5p 6p 7p 7p 9p 10p 12p 6p 7p 8p 8p 10p 12p IN4001 IN4002 IN4003 IN4004 1N4005 IN4006 100 200 400 600 IN4006 800 IN4007 1000

\$4-47

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When I read about the Signal Injector by Alan Jardine I was reminded of the poor beginner.

Following the instruction to solder the leads direct on to the cell will result in heating up the electrolyte and a very short life for the cell.

Perhaps this is not important as the choice of a push-on/push-off switch allows no easy means of knowing if the thing is on or off. Very few beginners will remember to test each time, and cell life will be short it is expected. A push button perhaps?

The blind cannot be expected to lead the blind, and beginners are usually short of experience.

R. Quorn Sussex

Of your two points concerning the Signal Injector, the first is a bit exaggerated. It is true that the cell life will be reduced by applying heat (from soldering iron) to the battery terminals but this is only negligible for the time required to execute the connection.

To install a holder to suit this type of battery would increase the cost by about 40 per cent.

We agree that it will be difficult to tell if the unit is on or off when not in use, but it can be determined; when the unit is "on" the push button will feel "loose" but in the "off" position this looseness disappears.

If this proves unsatisfactory a push-to-make release off type can be substituted.

More Accurate Timer

May I thank you for publishing another article combining the hobbies of electronics and photography (ref. *Darkroom Timer*, March issue).

Although of excellent design, I feel it must be stated that a timer with only a 5 second timing intervals is not nearly accurate enough for the demands of the high quality black and white or well balanced colour prints that are required. However, with a small modification, I have found that the timer may be converted to an accurate piece of equipment having a timing range of 5 to 45 seconds in one second steps.

The modification requires four extra components, which are a 5 position two-pole switch (S4), VR5, VR6 and VR7 which are

skeleton presets of the values, 5 kilohm, 10 kilohm and 20 kilohm respectively.

These components form an additional timing circuit which is connected in series with the original (R_t) .

Position 1 of the switch has no further resistance and acts as a short circuit; position 2 connects VR5 into circuit, whilst position 3 connects VR6; position 4 connects VR5 and VR6 and position 5 connects VR7 into circuit.

Each position of the new switch is to represent a further one second delay.

Position 1 of course, has no further delay, position 2 however, will give a one second delay, position 3 two seconds etc. when the presets are set as they were in the original timing circuit.

Now, any time, in one second steps may be selected from 5 to 45 seconds by selecting the required 5 second range, plus the required extra time (if any) on the new switch.

D. G. Smith Emsworth, Hants.

Components

Let me say first of all how much I enjoy your magazine and as a newcomer to electronics I find your *Teach-In* articles very interesting and also *Shop Talk*, etc. However, I wonder if I may make a suggestion?

I constructed your Demo-Deck and find that in following this series for a month or so there is a list of the more minor components used in the experiments and I wondered if it would be at all possible, either, preferably, if you could publish the list of all the components that would be required for the rest of this series in one complete list or if possible broken up into the individual months during which they will be required.

The reason I say this is, that I,

like many of your other readers no doubt, have no local supplier of components in my immediate vicinity and it usually means a trip to Edinburgh or Glasgow to purchase these components.

However, if I could have a full list this would make things much easier for me. It would also make it much easier to send off a full list by post to a mail order firm rather than asking for two or three small components every month or so. I wonder if this could be done.

I am very grateful to you and wish you every success for your future publications.

> R. L. Grant Scotland

It was our intention to publish an advanced list and in future we shall be publishing, at the end of each Teach-In every month, a list of components additional to those you have already acquired.

Calling Gloucester

Now that I'm receiving your magazine on regular order and greatly enjoying it, I feel that I ought to go a stage further in order to get any lasting benefit from your guidance.

Can I please find out through your pages how many people in the Gloucester area are willing to ask for, and attend, an evening class on useful, basic "everyday electronics"?

Should anyone be interested, could they please write to me at the address given, then provided enough wish it, our local Education Authority can be approached with evidence that the need for such a class does exist.

Many thanks for giving me a chance to ask for these people through your very sensible magazine.

E. L. Payn 82 Innsworth Lane, Longlevens, Gloucester GL2 0DE

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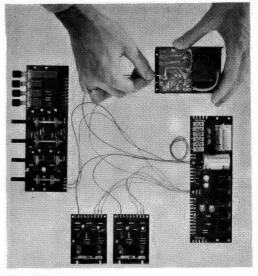
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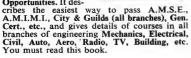
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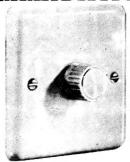
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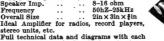
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